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The Teixeira singularity degeneracy and its bifurcation in PWL systems

Enrique Ponce



Joint work with Rony Cristiano & Daniel Pagano (UFSC, Florianópolis, Brasil) and Emilio Freire (Univ. de Sevilla),



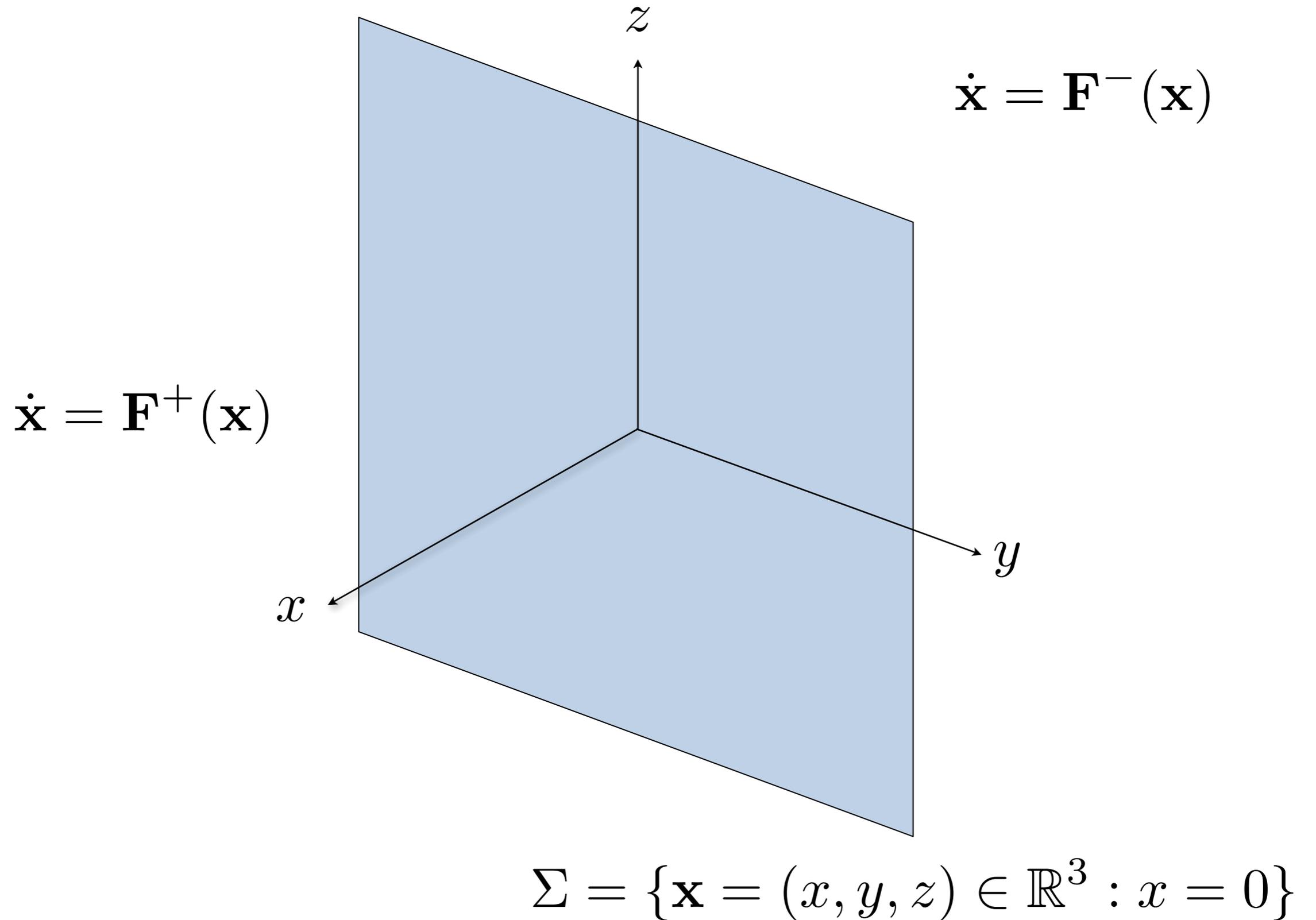
Introduction

- We consider 3D piecewise linear Filippov differential systems with a separation plane, having a two-fold point with invisible tangencies, that is, the so-called Teixeira singularity (TS-point, for short).
- For some parameter values this singularity undergoes a compound bifurcation: there appears a sliding bifurcation involving a pseudo-equilibrium point and, simultaneously, a bifurcation associated to the birth of a crossing limit cycle.
- After determining a generic canonical form, we show how to characterize such a compound bifurcation.
- Our motivation comes from the natural appearance of TS-points in the control of Boost converters.

Summary

- A non-trivial yet manageable single-parameter example
- The general DPWL case:
 - Hypotheses for having a TS-point and canonical form
 - The sliding dynamics bifurcation
 - The crossing dynamics bifurcation
- A multi-parametric example
- TS-point in a DC-DC Boost converter
- Conclusions

A non-trivial yet manageable example

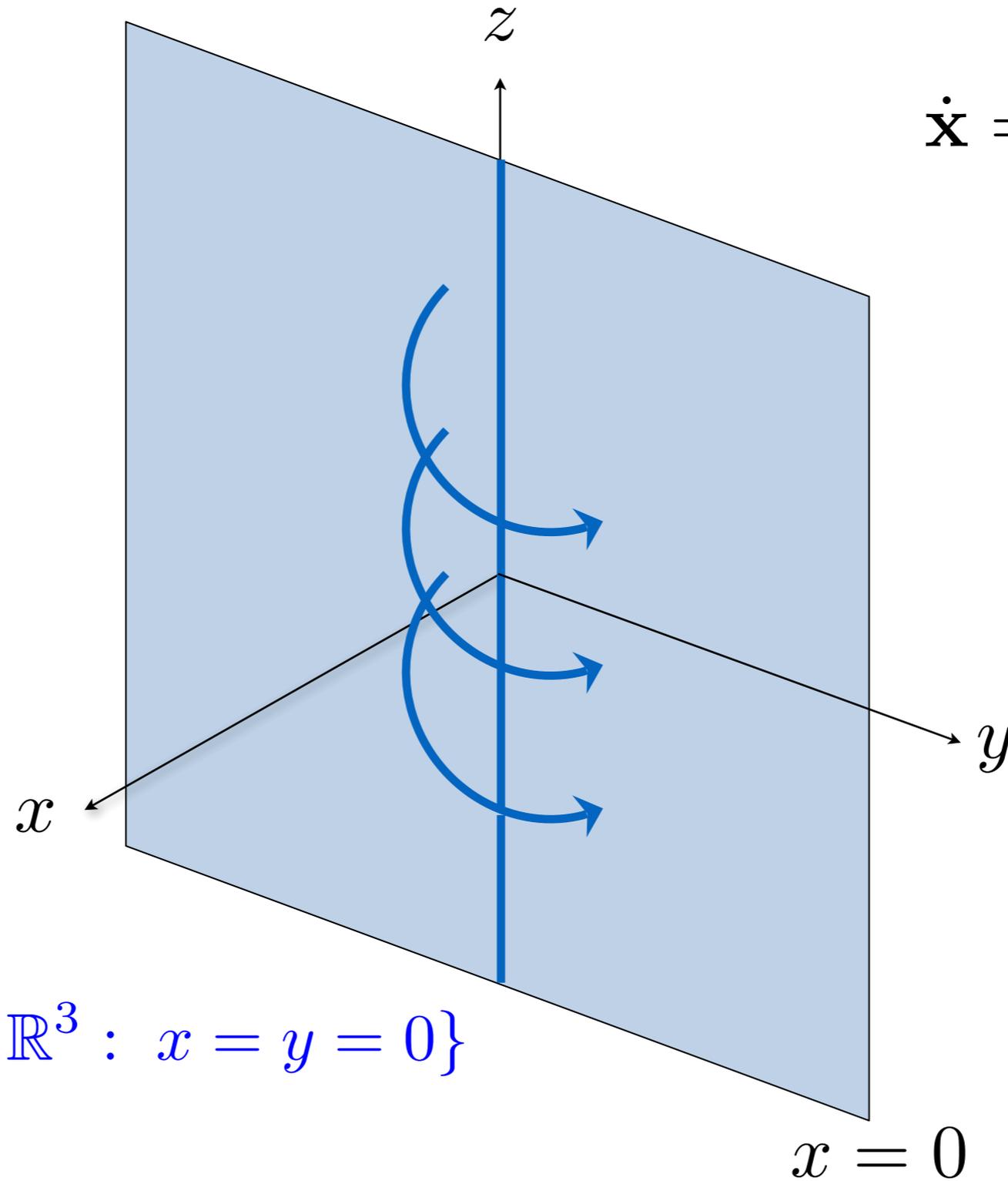


A non-trivial yet manageable example

$$\ddot{x} = -\dot{y} = -1$$

$$\begin{cases} \dot{x} = -y \\ \dot{y} = 1 \\ \dot{z} = -2 \end{cases}$$

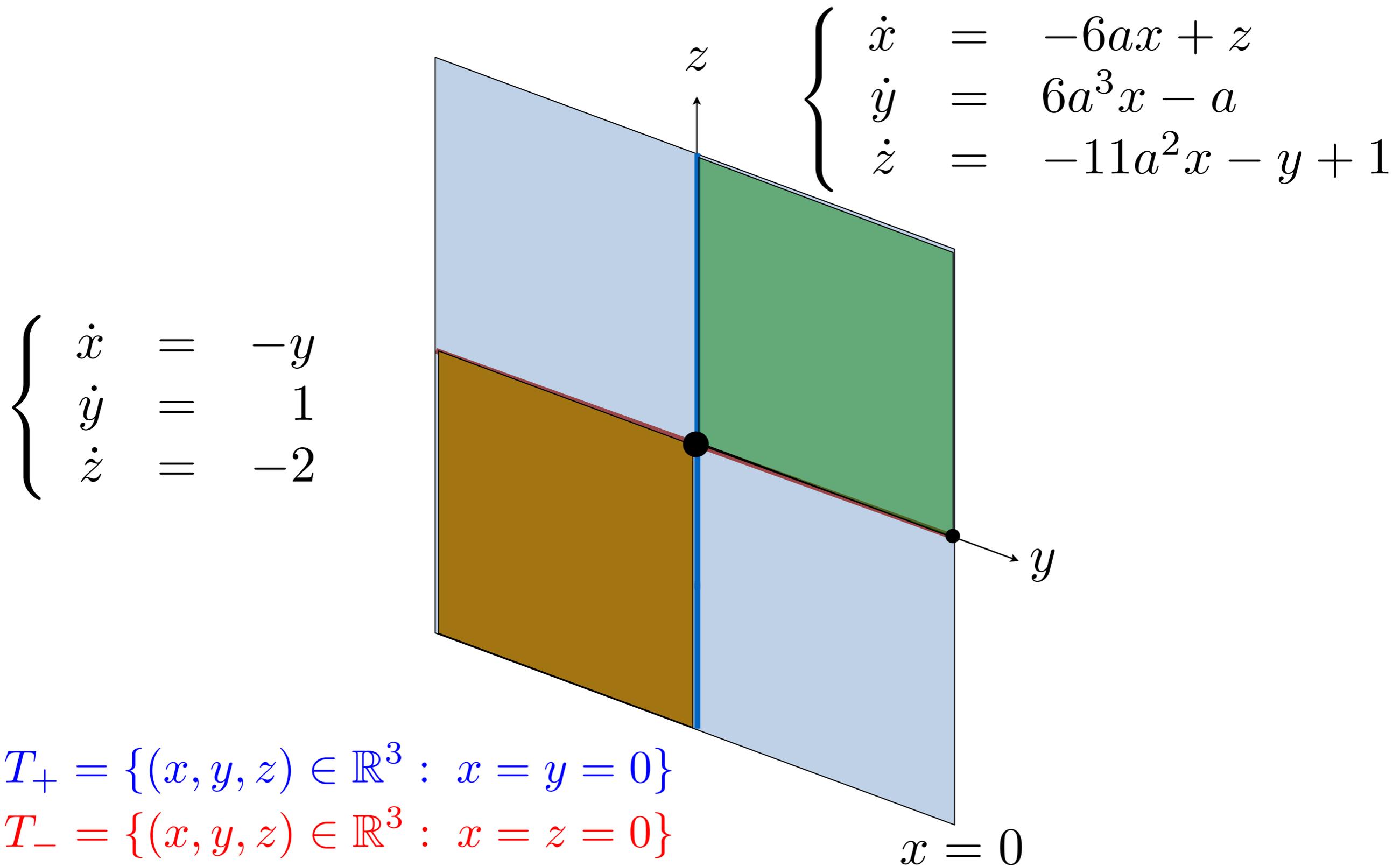
$$\dot{\mathbf{x}} = \mathbf{F}^{-}(\mathbf{x})$$



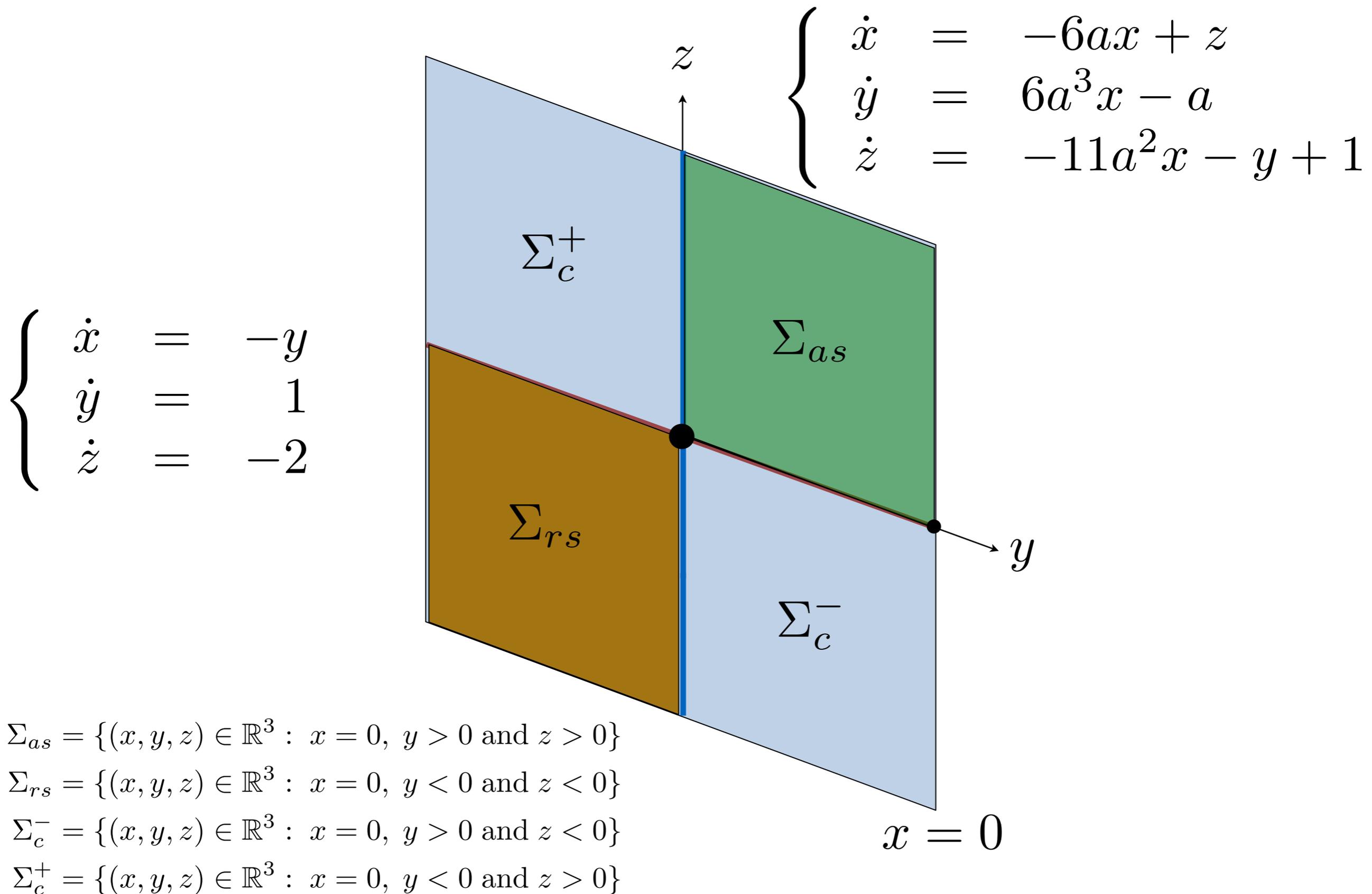
$$T_+ = \{(x, y, z) \in \mathbb{R}^3 : x = y = 0\}$$

$$x = 0$$

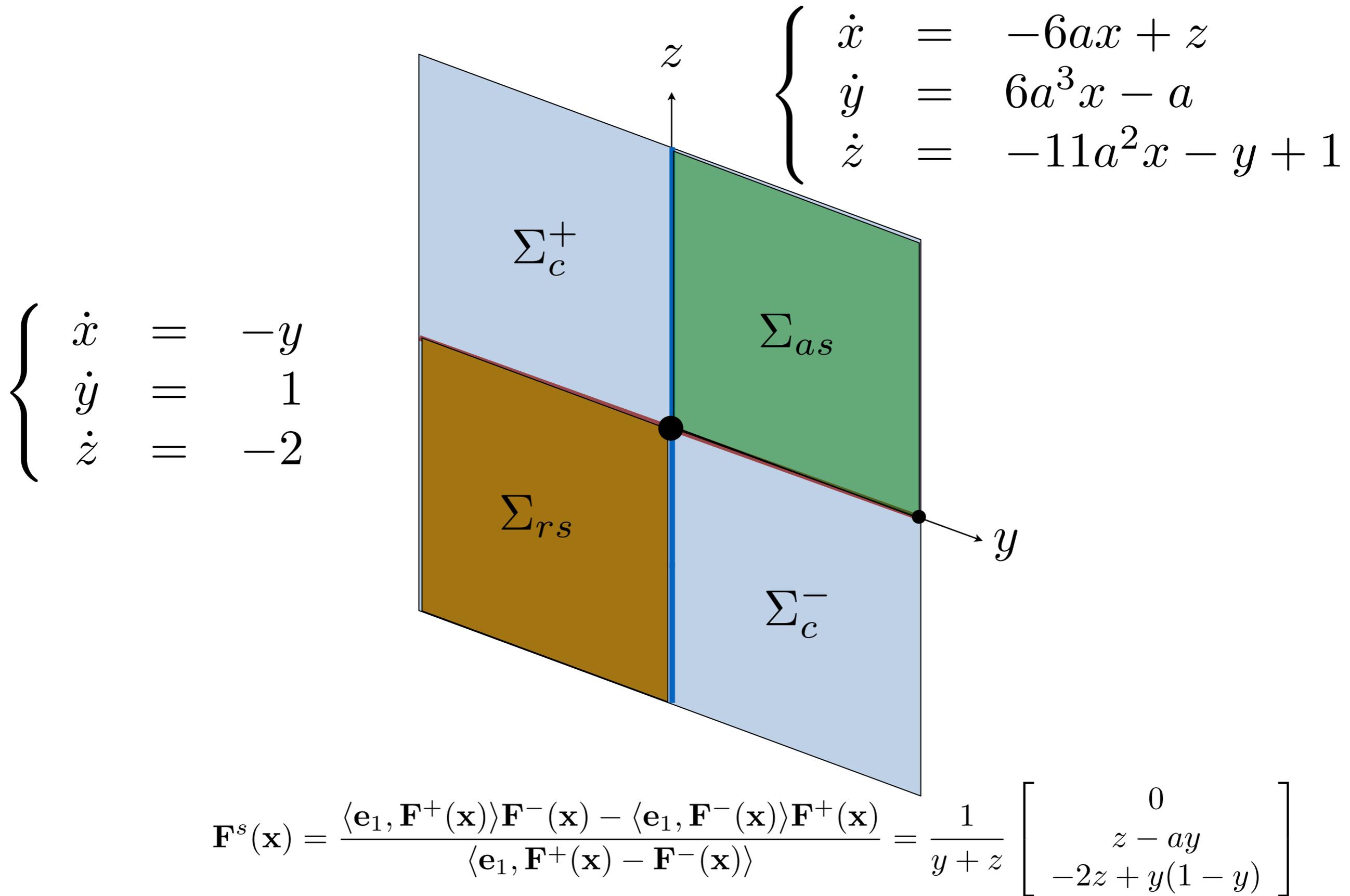
A non-trivial yet manageable example



A non-trivial yet manageable example



The sliding dynamics



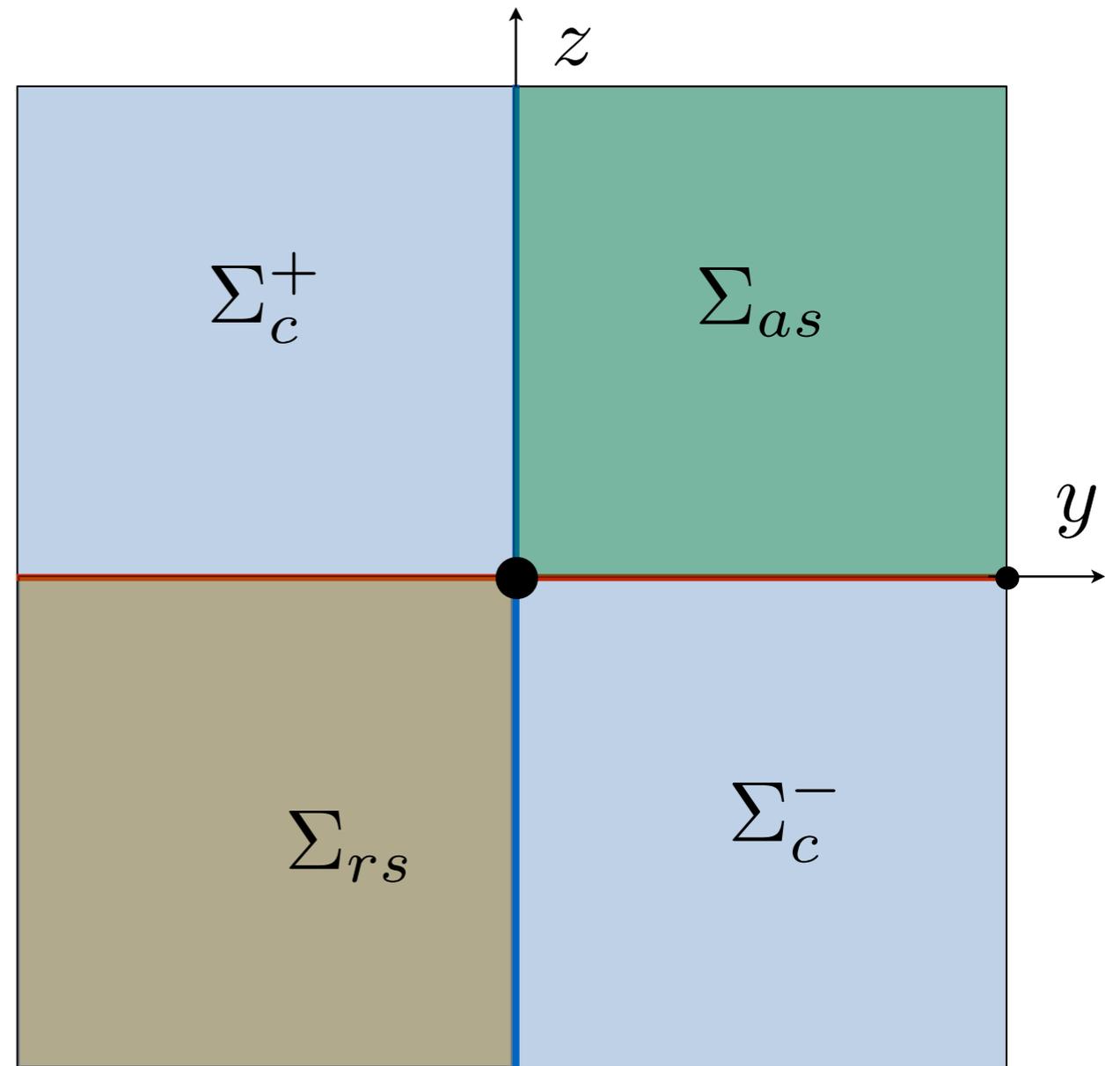
The sliding dynamics

$$\mathbf{F}^s(\mathbf{x}) = \frac{1}{y+z} \begin{bmatrix} 0 \\ z - ay \\ -2z + y(1-y) \end{bmatrix}$$



Pseudo-equilibrium point at

$$\mathbf{x}(a) = (0, y(a), z(a)) = (0, 1 - 2a, a(1 - 2a))$$



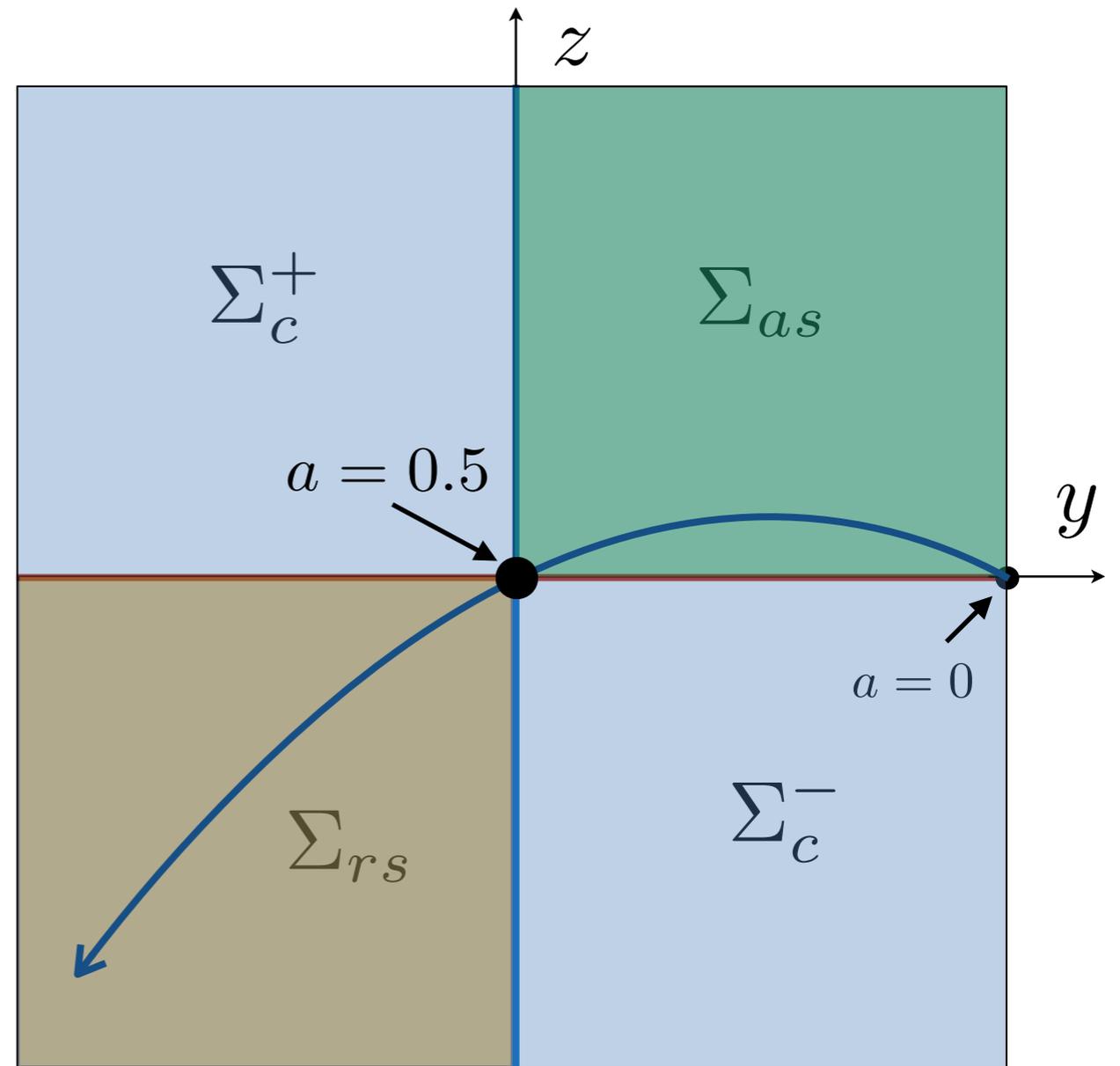
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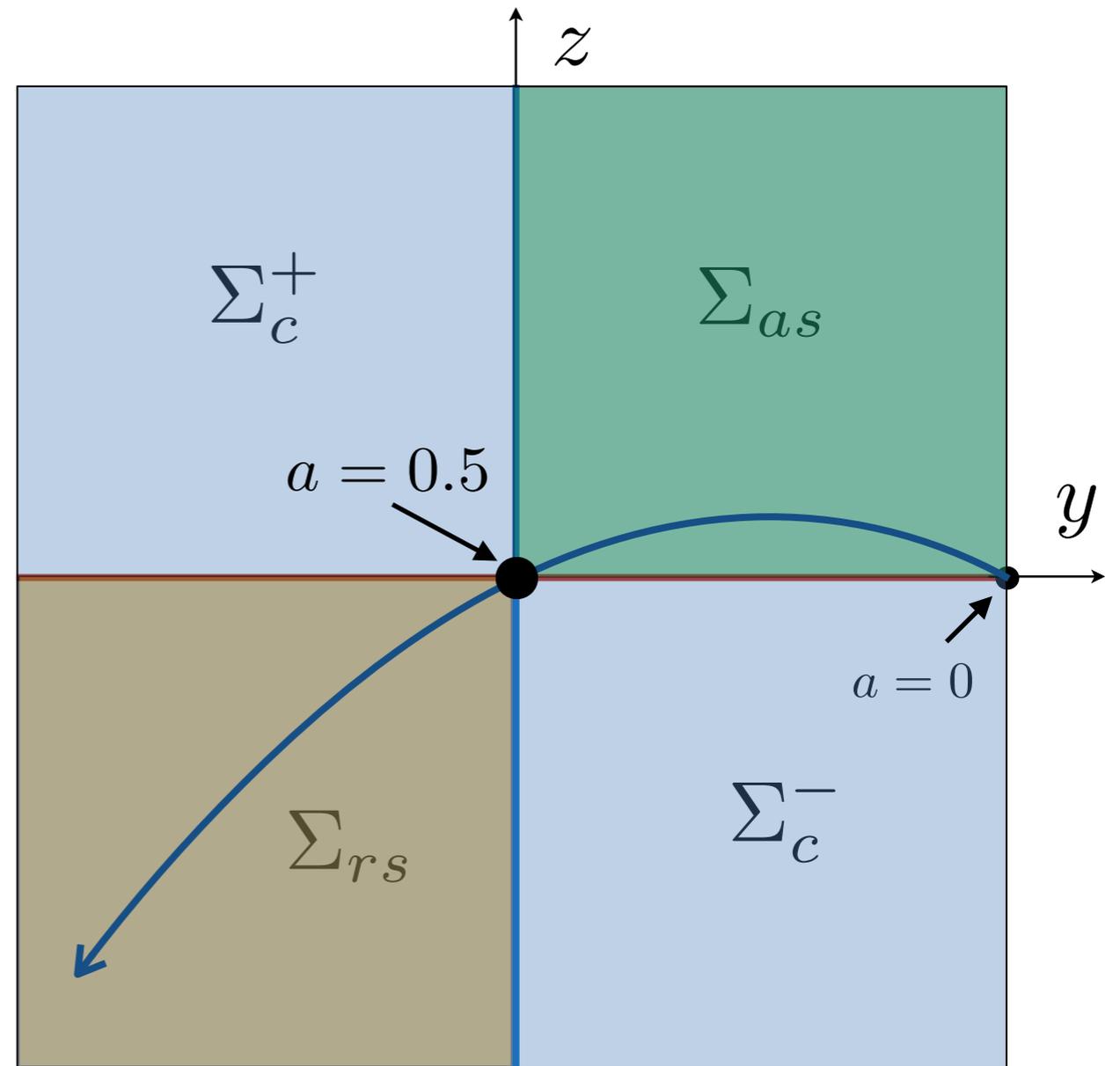
The sliding dynamics

Desingularized sliding v.f.

$$\mathbf{F}_d^s(\mathbf{x}) = \pm \begin{bmatrix} 0 \\ z - ay \\ -2z + y(1 - y) \end{bmatrix}$$

$$\pm(y + z) > 0$$

$$\mathbf{F}^s(\mathbf{x}) = \frac{1}{y + z} \begin{bmatrix} 0 \\ z - ay \\ -2z + y(1 - y) \end{bmatrix}$$



Pseudo-equilibrium point at

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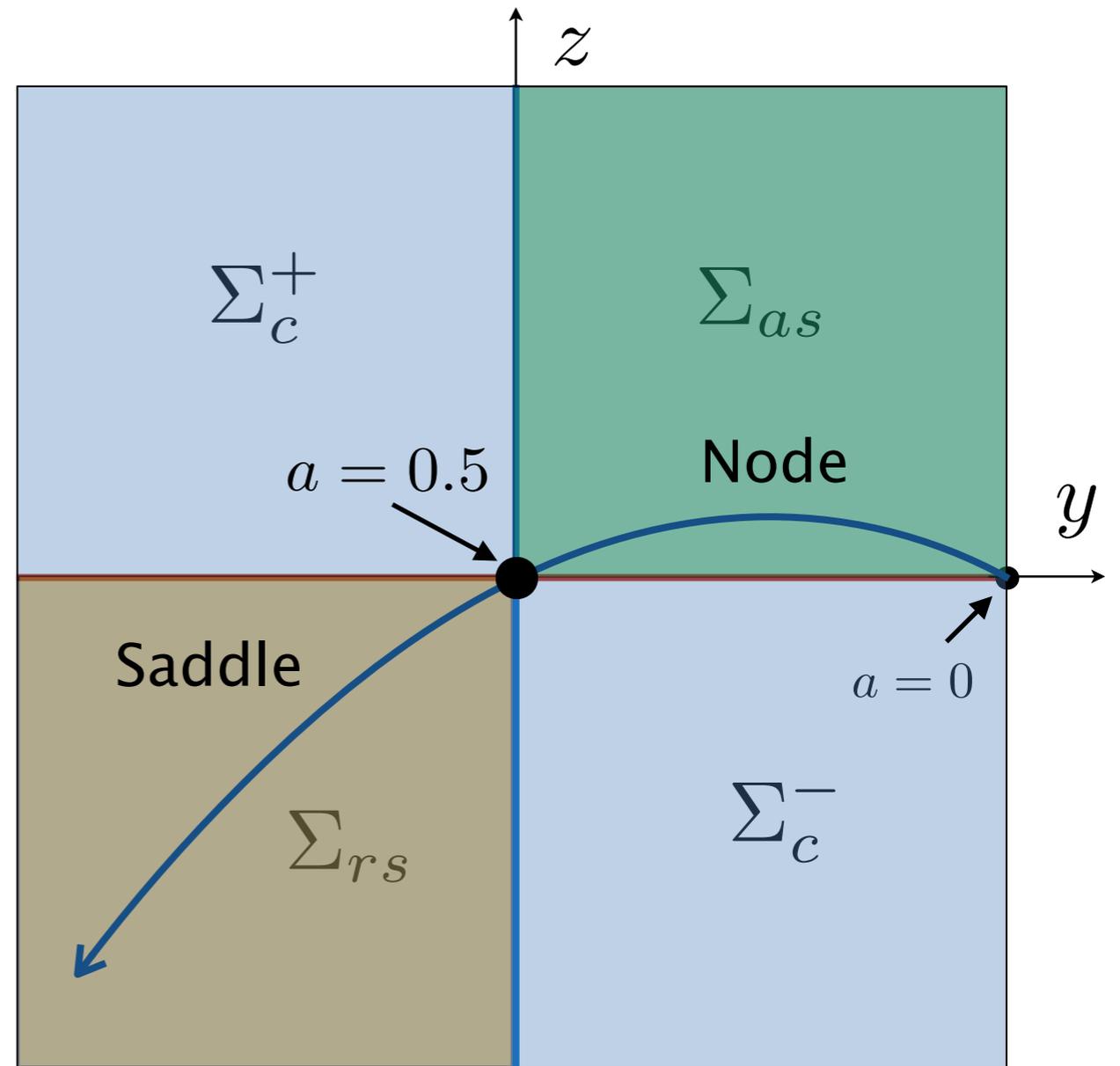
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Pseudo-equilibrium point at

$$\mathbf{x}(a) = (0, y(a), z(a)) = (0, 1 - 2a, a(1 - 2a))$$

The crossing dynamics

The integration of the (-) vector field is much more involved but the choice of stepped eigenvalues allows to express solutions in an algebraic way.

$$A^- = \begin{pmatrix} -6a & 0 & 1 \\ 6a^3 & 0 & 0 \\ -11a^2 & -1 & 0 \end{pmatrix} \quad \text{Spec}(A^-) = \{-a, -2a, -3a\}$$

$$e^{-A^-t} = \frac{e^{at}}{2} \begin{pmatrix} 1 - 8e^{at} + 9e^{2at} & -\frac{(-1+e^{at})^2}{a^2} & -\frac{(-1+e^{at})(-1+3e^{at})}{a} \\ -6a^2(-1+e^{at})(-1+3e^{at}) & 2(3-3e^{at}+e^{2at}) & 6a(-1+e^{at})^2 \\ a(-1+e^{at})(-5+27e^{at}) & -\frac{(-1+e^{at})(-5+3e^{at})}{a} & -5+16e^{at}-9e^{2at} \end{pmatrix} =$$

$$= \frac{1}{2u^3} \begin{pmatrix} u^2 - 8u + 9 & -\frac{(u-1)^2}{a^2} & -\frac{(u-3)(u-1)}{a} \\ -6a^2(u-3)(u-1) & 2(3u^2 - 3u + 1) & 6a(u-1)^2 \\ a(u-1)(5u-27) & -\frac{(u-1)(5u-3)}{a} & -5u^2 + 16u - 9 \end{pmatrix}$$

$$0 < u = e^{-at} < 1, \quad a > 0$$

Looking for $P_-^{-1}(y, z)$ we write

$$\begin{bmatrix} x_-^{-1}(t) \\ y_-^{-1}(t) \\ z_-^{-1}(t) \end{bmatrix} = \frac{1}{6a^2} \begin{bmatrix} 1 \\ -5a^2 \\ 6a \end{bmatrix} + e^{-A^-t} \left(\begin{bmatrix} 0 \\ y_0 \\ z_0 \end{bmatrix} - \frac{1}{6a^2} \begin{bmatrix} 1 \\ -5a^2 \\ 6a \end{bmatrix} \right)$$

The condition $x_-^{-1}(t) = 0$ gives

$$0 < u = e^{-at} < 1, \quad a > 0$$

$$a = \frac{(-1 + u)(2 + u - 3y_0)}{3(-3 + u)z_0}, \quad \text{and then}$$

$$y_-^{-1}(u) = \frac{-1 - 9u + 9u^2 + u^3 + 6(1 - 3u)y_0}{6(-3 + u)u^2},$$

$$z_-^{-1}(u) = \frac{(-1 - 6u + u^2 + 6y_0)z_0}{2u^2(2 + u - 3y_0)}$$

Finally, imposing

$$P_+ \begin{pmatrix} y_0 \\ z_0 \end{pmatrix} = \begin{bmatrix} -2y_0 \\ -4y_0 + z_0 \end{bmatrix} = P_-^{-1} \begin{pmatrix} y_0 \\ z_0 \end{pmatrix} = \begin{bmatrix} y_-^{-1}(u) \\ z_-^{-1}(u) \end{bmatrix}$$

we get

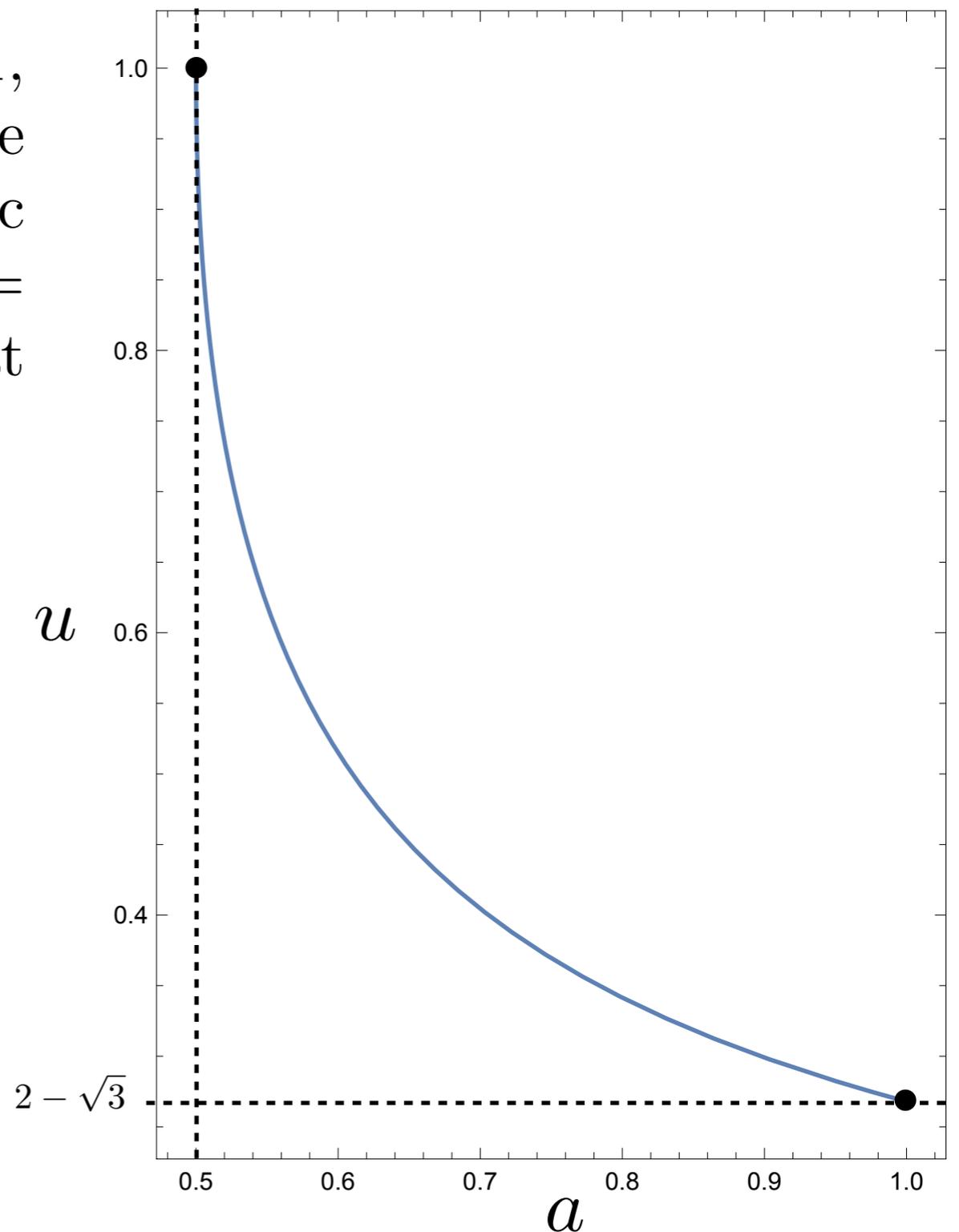
$$a = \frac{1 + 2u + 6u^2 + 2u^3 + u^4}{2u(1 + 10u + u^2)}, \quad \text{and then}$$

$$y_0 = -\frac{(-1 + u)(1 + 10u + u^2)}{6(1 + u)(1 - 4u + u^2)},$$

$$z_0 = \frac{(-1 + u)(-1 + 6u + 5u^2 + 2u^3)}{6a(1 + u)(1 - 4u + u^2)}$$

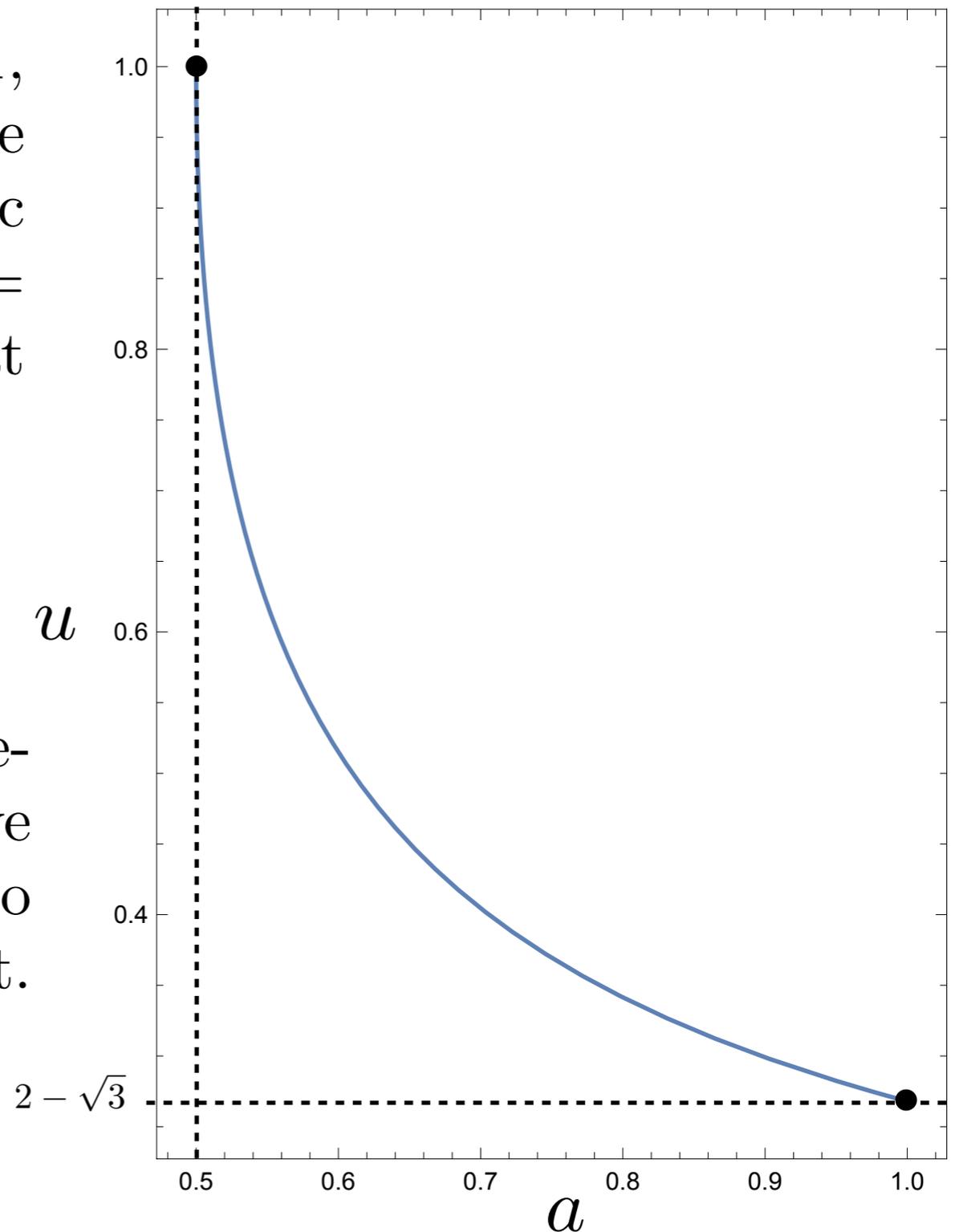
Note that $1 - 4u + u^2 = (u - 2)^2 - 3$, so that the above expressions are only valid for $2 - \sqrt{3} < u < 1$. Also, $\lim_{u \rightarrow 1^-} a = 1/2$, $\lim_{u \rightarrow (2 - \sqrt{3})^+} a = \infty$.

A periodic orbit exists for $0.5 < a < 1$, corresponding to a decreasing of u in the interval $2 - \sqrt{3} < u < 1$. The periodic orbit is born from the TS-point at $a = 0.5$ and disappears in a bifurcation at infinity for $a = 1$.

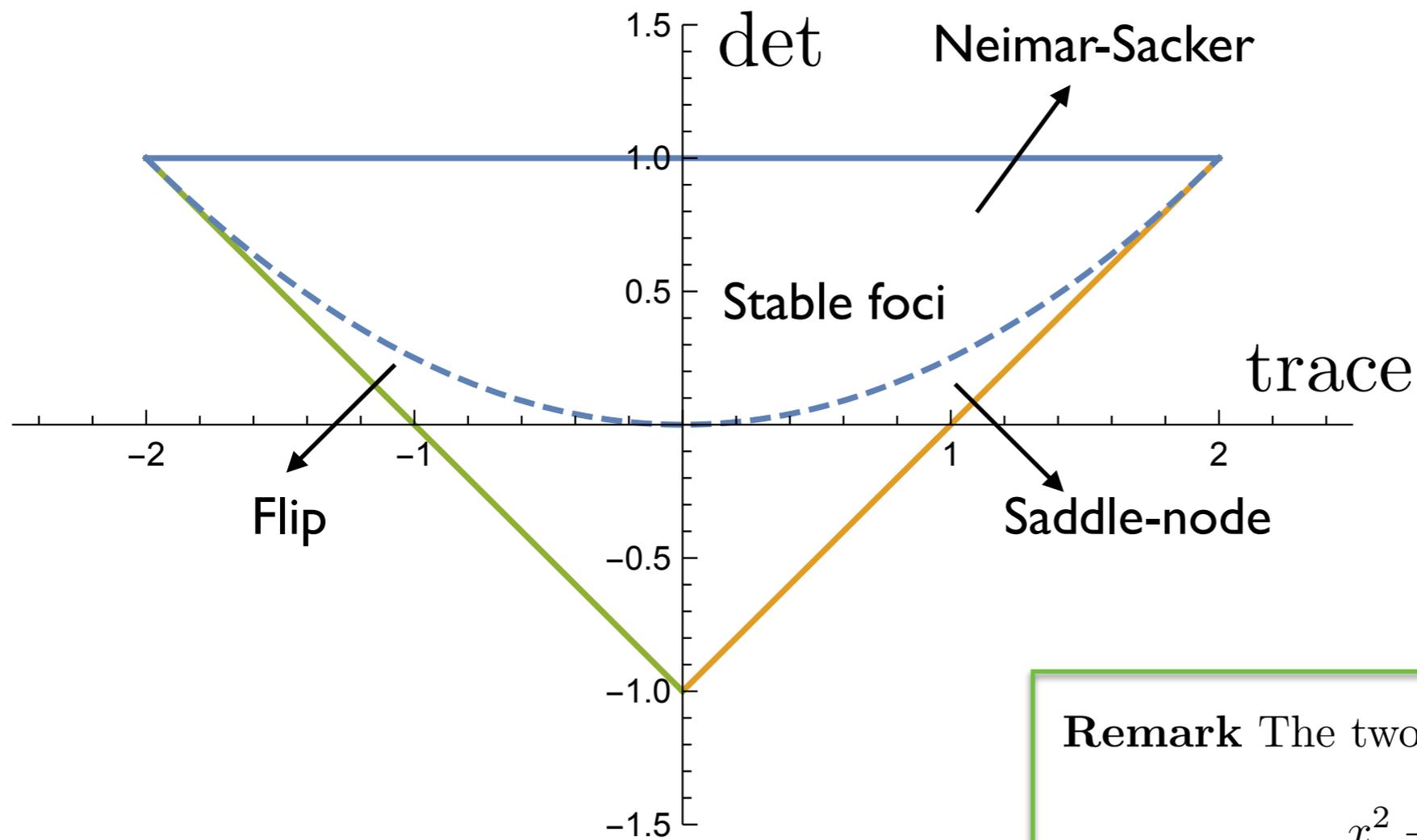


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We can compute in terms of u , the determinant and trace of the derivative $DP(y_0, z_0) = DP_-^{-1} \circ DP_+(y_0, z_0)$ to check the stability of the periodic orbit.



The stability triangle in the (trace,det)-plane



Remark The two roots of the quadratic equation

$$x^2 - \text{trace} \cdot x + \text{det} = 0$$

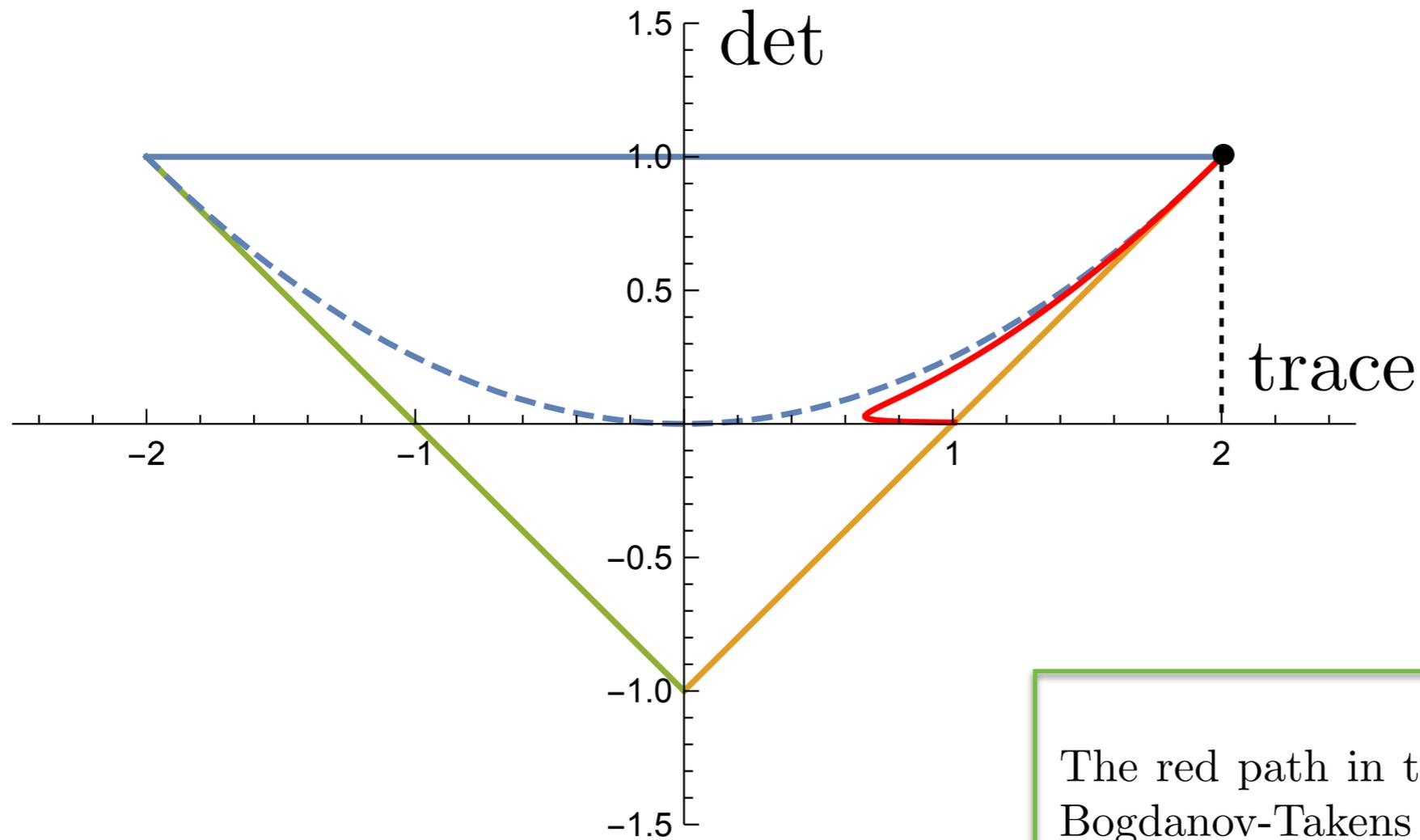
are in the interior of the unit disk of the complex plane if and only if the two inequalities

$$|\text{det}| < 1$$

$$|\text{trace}| < \text{det} + 1$$

hold.

The followed path for $0.5 < a < 1$ in the (trace,det)-plane



The red path in the (trace,det)-plane starts at the Bogdanov-Takens point $(2,1)$. Note that the final point in the path is over the horizontal axis, namely

$$(98 - 56\sqrt{3}, 97 - 56\sqrt{3}).$$

For this example, the crossing periodic orbit which is born from the TS-point is always of **stable node** type.

Such a global control of the compound bifurcation for the Teixeira singularity is rather unusual: in what follows, we show how it is possible to characterize locally this bifurcation for discontinuous linear systems.

References

(*) A. F. Filippov, Differential Equations with Discontinuous Righthand Sides, Kluwer Academic Publishers, Dordrecht, 1988.

(*) M.A. Teixeira, Stability conditions for discontinuous vector fields, Journal of Differential Equations 88 (1990) 15–29.

(*) M.A. Teixeira, Generic bifurcation of sliding vector fields, J. Math. Anal. Appl. 176 (1993) 436–457.

(*) M. R. Jeffrey, A. Colombo, The two-fold singularity of discontinuous vector fields, SIAM J. Applied Dynamical Systems 8 (2) (2009) 624–640.

(*) A. Colombo, M. di Bernardo, E. Fossas, M. R. Jeffrey, Teixeira singularities in 3d switched feedback control systems, Systems & Control Letters 59 (2010) 615–622.

(*) A. Colombo, M. R. Jeffrey, Nondeterministic chaos, and the two-fold singularity in piece-wise smooth flows, SIAM J. Applied Dynamical Systems 10 (2) (2011) 423–451.

(*) R. Cristiano, D. J. Pagano, L. Benadero, and E. Ponce, Bifurcation analysis of a DC-DC bidirectional power converter operating with constant power load, International Journal of Bifurcation and Chaos, 26 (2016), p. 1630010

(*) E. Freire, R. Cristiano, D. Pagano and E. Ponce, Bifurcation analysis of the Teixeira singularity in three-dimensional piecewise linear dynamical systems, preprint 2016.

The general DPWL case

There exist three vectors \mathbf{w}^- , \mathbf{w}^+ , $\mathbf{v} \in \mathbb{R}^3$, with $\mathbf{v} \neq \mathbf{0}$, two 3×3 square matrices A^- and A^+ and a scalar $\delta \in \mathbb{R}$, such that

$$(a) \quad \mathbf{F}(\mathbf{x}) = \begin{cases} \mathbf{F}^-(\mathbf{x}) = A^- \mathbf{x} + \mathbf{w}^-, & \text{if } \mathbf{v}^T \mathbf{x} + \delta < 0 \\ \mathbf{F}^+(\mathbf{x}) = A^+ \mathbf{x} + \mathbf{w}^+, & \text{if } \mathbf{v}^T \mathbf{x} + \delta > 0 \end{cases}$$

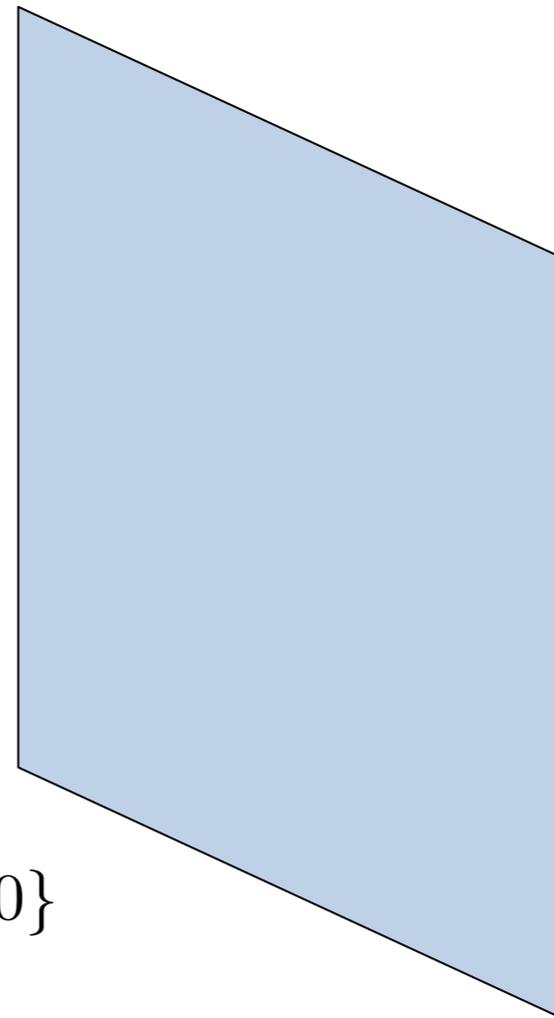
(b) $A^- \mathbf{x} + \mathbf{w}^- \neq A^+ \mathbf{x} + \mathbf{w}^+$ generically, when $\mathbf{v}^T \mathbf{x} + \delta = 0$.

$$\dot{\mathbf{x}} = \mathbf{F}^+(\mathbf{x})$$

$$\dot{\mathbf{x}} = \mathbf{F}^-(\mathbf{x})$$

The switching manifold is the plane

$$\Sigma = \{\mathbf{x} \in \mathbb{R}^3 : \mathbf{v}^T \mathbf{x} + \delta = 0\}$$



Hypotheses for having a TS-point

(H1) The tangency lines

$$T^- = \{\mathbf{x} \in \Sigma : \langle \mathbf{v}, \mathbf{F}^-(\mathbf{x}) \rangle = 0\}$$

$$T^+ = \{\mathbf{x} \in \Sigma : \langle \mathbf{v}, \mathbf{F}^+(\mathbf{x}) \rangle = 0\}$$

intersect transversally at a point $\hat{\mathbf{x}}$.

(H2) At the two-fold point $\hat{\mathbf{x}}$ both tangencies are invisible: $\mathbf{v}^T A^- \mathbf{F}^-(\hat{\mathbf{x}}) > 0$ and $\mathbf{v}^T A^+ \mathbf{F}^+(\hat{\mathbf{x}}) < 0$.

The canonical form for DPWLS with a TS-point

Proposition. Under hypotheses **(H1)** and **(H2)**, it is possible through an invertible linear change of coordinates and a rescaling of time, to rewrite the system in the canonical form

$$\dot{\mathbf{x}} = \begin{cases} \mathbf{F}^-(\mathbf{x}), & \text{if } x < 0, \\ \mathbf{F}^+(\mathbf{x}), & \text{if } x > 0, \end{cases} \quad (1)$$

with $\mathbf{x} = (x, y, z) \in \mathbb{R}^3$, where the linear vector fields $\mathbf{F}^\pm : \mathbb{R}^3 \rightarrow \mathbb{R}^3$ are

$$\mathbf{F}^+(\mathbf{x}) = \begin{bmatrix} f_1^+ & -1 & 0 \\ g_1^+ & g_2^+ & g_3^+ \\ h_1^+ & h_2^+ & h_3^+ \end{bmatrix} \begin{bmatrix} x \\ y \\ z \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \\ v_+ \end{bmatrix}, \quad \mathbf{F}^-(\mathbf{x}) = \begin{bmatrix} f_1^- & 0 & 1 \\ g_1^- & g_2^- & g_3^- \\ h_1^- & h_2^- & h_3^- \end{bmatrix} \begin{bmatrix} x \\ y \\ z \end{bmatrix} + \begin{bmatrix} 0 \\ v_- \\ 1 \end{bmatrix},$$

for some constants $f_1^\pm, g_1^\pm, g_2^\pm, g_3^\pm, h_1^\pm, h_2^\pm, h_3^\pm$ and v_\pm .

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for some constants $f_1^\pm, g_1^\pm, g_2^\pm, g_3^\pm, h_1^\pm, h_2^\pm, h_3^\pm$ and v_\pm .

$$T_+ = \{(x, y, z) \in \mathbb{R}^3 : x = y = 0\}$$

$$T_- = \{(x, y, z) \in \mathbb{R}^3 : x = z = 0\}$$

$$\Sigma = \{\mathbf{x} = (x, y, z) \in \mathbb{R}^3 : x = 0\}$$

TS-point at the origin!

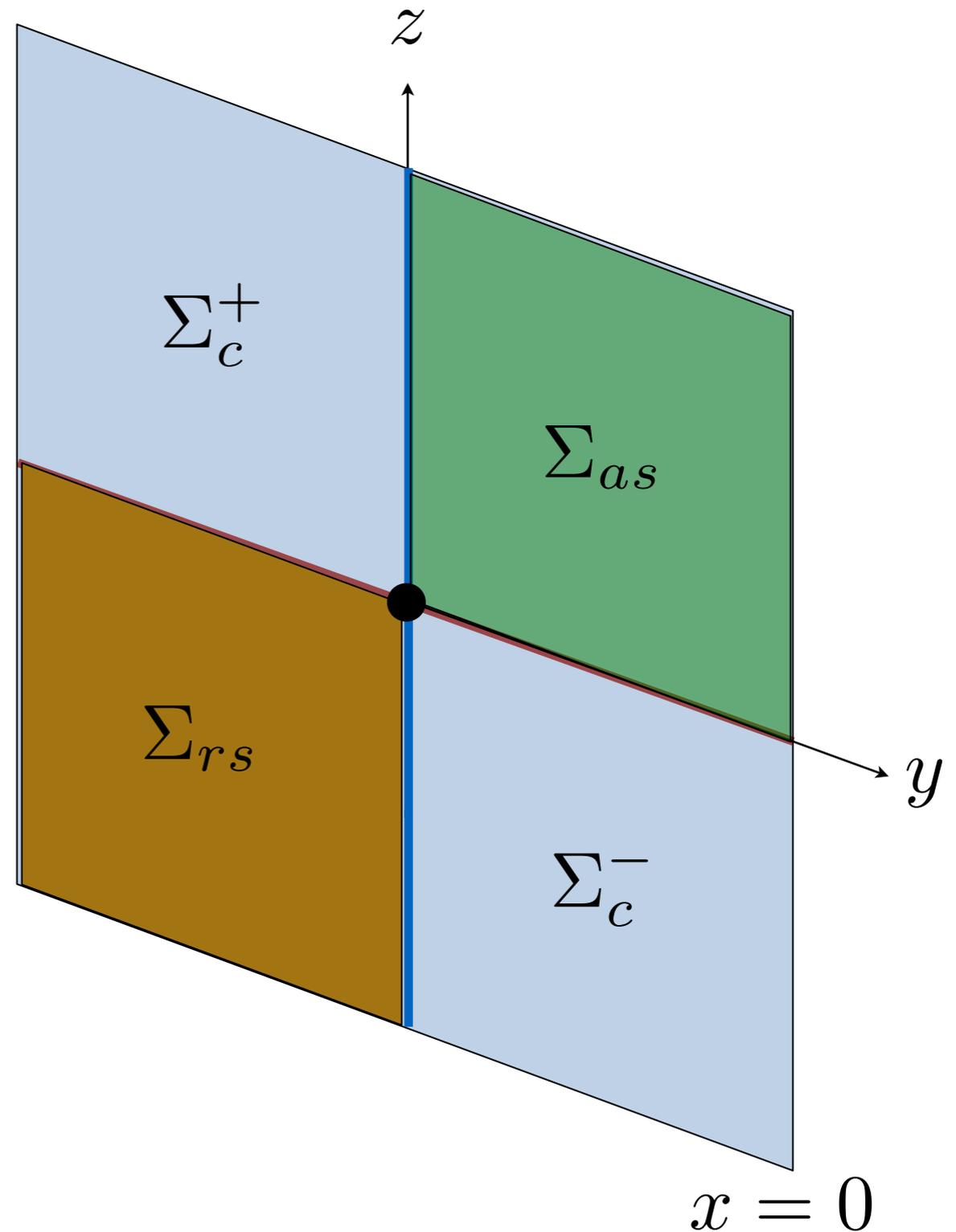
The canonical form for DPWLS with a TS-point

$$\Sigma_{as} = \{(x, y, z) \in \mathbb{R}^3 : x = 0, y > 0 \text{ and } z > 0\}$$

$$\Sigma_{rs} = \{(x, y, z) \in \mathbb{R}^3 : x = 0, y < 0 \text{ and } z < 0\}$$

$$\Sigma_c^- = \{(x, y, z) \in \mathbb{R}^3 : x = 0, y > 0 \text{ and } z < 0\}$$

$$\Sigma_c^+ = \{(x, y, z) \in \mathbb{R}^3 : x = 0, y < 0 \text{ and } z > 0\}$$



The sliding dynamics bifurcation

The sliding vector field associated to system (1) is

$$\mathbf{F}^s(0, y, z) = \frac{1}{y+z} \begin{bmatrix} 0 \\ v_- y + z + g_2^- y^2 + (g_2^+ + g_3^-)yz + g_3^+ z^2 \\ y + v_+ z + h_2^- y^2 + (h_2^+ + h_3^-)yz + h_3^+ z^2 \end{bmatrix}.$$

We work with the desingularized system

$$\dot{\mathbf{x}} = \mathbf{F}_d^s(0, y, z) := \begin{bmatrix} 0 \\ v_- y + z + g_2^- y^2 + (g_2^+ + g_3^-)yz + g_3^+ z^2 \\ y + v_+ z + h_2^- y^2 + (h_2^+ + h_3^-)yz + h_3^+ z^2 \end{bmatrix} \quad (2)$$

The sliding dynamics bifurcation

Pseudo-equilibrium points come from solving the two equations

$$v_- y + z + [y \quad z] G \begin{bmatrix} y \\ z \end{bmatrix} = 0,$$

$$y + v_+ z + [y \quad z] H \begin{bmatrix} y \\ z \end{bmatrix} = 0,$$

where

$$G = \begin{bmatrix} g_2^- & g_3^- \\ g_2^+ & g_3^+ \end{bmatrix} \quad \text{and} \quad H = \begin{bmatrix} h_2^- & h_3^- \\ h_2^+ & h_3^+ \end{bmatrix}.$$

The sliding dynamics bifurcation

Pseudo-equilibrium points come from solving the two equations

$$J(0,0) = \begin{bmatrix} v_- & 1 \\ 1 & v_+ \end{bmatrix} \leftarrow \begin{cases} v_- y + z + [y \ z] G \begin{bmatrix} y \\ z \end{bmatrix} = 0, \\ y + v_+ z + [y \ z] H \begin{bmatrix} y \\ z \end{bmatrix} = 0, \end{cases}$$

where

$$G = \begin{bmatrix} g_2^- & g_3^- \\ g_2^+ & g_3^+ \end{bmatrix} \quad \text{and} \quad H = \begin{bmatrix} h_2^- & h_3^- \\ h_2^+ & h_3^+ \end{bmatrix}.$$

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where

$$G = \begin{bmatrix} g_2^- & g_3^- \\ g_2^+ & g_3^+ \end{bmatrix} \quad \text{and} \quad H = \begin{bmatrix} h_2^- & h_3^- \\ h_2^+ & h_3^+ \end{bmatrix}.$$

Apart from the origin, we can have another non-trivial emanating branch for $v_- v_+ = 1$ (transcritical bifurcation). We assume all parameters fixed excepting

$$v_- = v_-(\varepsilon) := \frac{1 + \varepsilon}{v_+}$$

The sliding dynamics bifurcation

Pseudo-equilibrium points come from solving the two equations

$$J(0, 0) = \begin{bmatrix} v_- & 1 \\ 1 & v_+ \end{bmatrix} \leftarrow \begin{cases} v_- y + z + [y \ z] G \begin{bmatrix} y \\ z \end{bmatrix} = 0, \\ y + v_+ z + [y \ z] H \begin{bmatrix} y \\ z \end{bmatrix} = 0, \end{cases}$$

where

$$G = \begin{bmatrix} g_2^- & g_3^- \\ g_2^+ & g_3^+ \end{bmatrix} \quad \text{and} \quad H = \begin{bmatrix} h_2^- & h_3^- \\ h_2^+ & h_3^+ \end{bmatrix}.$$

$$\underbrace{(v_+ v_- - 1)}_{\varepsilon} y + [y \ z] (v_+ G - H) \begin{bmatrix} y \\ z \end{bmatrix} = 0,$$

$$y + v_+ z + [y \ z] H \begin{bmatrix} y \\ z \end{bmatrix} = 0.$$

The sliding dynamics bifurcation

We can use the equivalent system of equations

$$\varepsilon y + [y \quad z] (v_+ G - H) \begin{bmatrix} y \\ z \end{bmatrix} = 0,$$
$$y + v_+ z + [y \quad z] H \begin{bmatrix} y \\ z \end{bmatrix} = 0.$$

Using the implicit function theorem for the second equation at $(y, z) = (0, 0)$ we get that, for z small, solutions must satisfy

$$\begin{bmatrix} y \\ z \end{bmatrix} = \begin{bmatrix} -v_+ \\ 1 \end{bmatrix} z + O(z^2).$$

We need $v_+ < 0$

Substituting this expansion in the first equation, desingularizing it and assuming

$$\kappa_S = [-v_+ \quad 1] (v_+ G - H) \begin{bmatrix} -v_+ \\ 1 \end{bmatrix} \neq 0,$$

we get

$$z = \frac{v_+}{\kappa_S} \varepsilon + O(\varepsilon^2).$$

The sliding dynamics bifurcation

Proposition (Pseudo-equilibrium transition) Assuming that the criticality coefficient

$$\kappa_S = \begin{bmatrix} -v_+ & 1 \end{bmatrix} (v_+G - H) \begin{bmatrix} -v_+ \\ 1 \end{bmatrix} \neq 0$$

the following statements hold.

- (a) System (2) undergoes a transcritical bifurcation for $\varepsilon = 0$, so that there exists a branch of equilibria $(\tilde{y}(\varepsilon), \tilde{z}(\varepsilon))$ with $(\tilde{y}(0), \tilde{z}(0)) = (0, 0)$ and $(\tilde{y}'(0), \tilde{z}'(0)) = \left(\frac{-v_+^2}{\kappa_S}, \frac{v_+}{\kappa_S} \right)$.
- (b) For the particular case where $v_+ < 0$, the emanating branch is located at the quadrants with $yz > 0$. If $\kappa_S > 0$ ($\kappa_S < 0$) then in passing from $\varepsilon < 0$ to $\varepsilon > 0$ the origin passes from being a saddle to a stable node, while the nontrivial equilibrium passes from being a stable node in the first (third) quadrant to be a saddle in the third (first) quadrant.

The sliding dynamics bifurcation

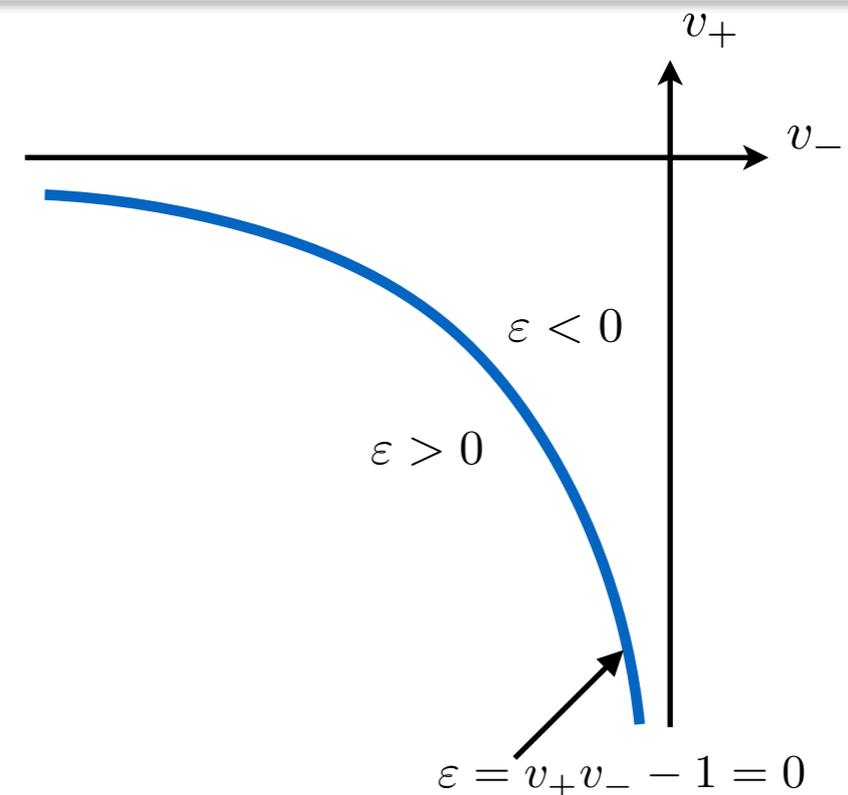
Theorem S Assuming $v_+ < 0$, $v_- < 0$ and $\kappa_S \neq 0$, system (1) has for $\varepsilon = v_-v_+ - 1$ with $|\varepsilon| > 0$ small, one pseudo-equilibrium point $\tilde{\mathbf{x}}(\varepsilon) = (0, \tilde{y}(\varepsilon), \tilde{z}(\varepsilon))$, such that

$$(\tilde{y}(\varepsilon), \tilde{z}(\varepsilon)) = \left(-\frac{v_+^2}{\kappa_S}, \frac{v_+}{\kappa_S} \right) \varepsilon + O(\varepsilon^2),$$

and the following statements hold.

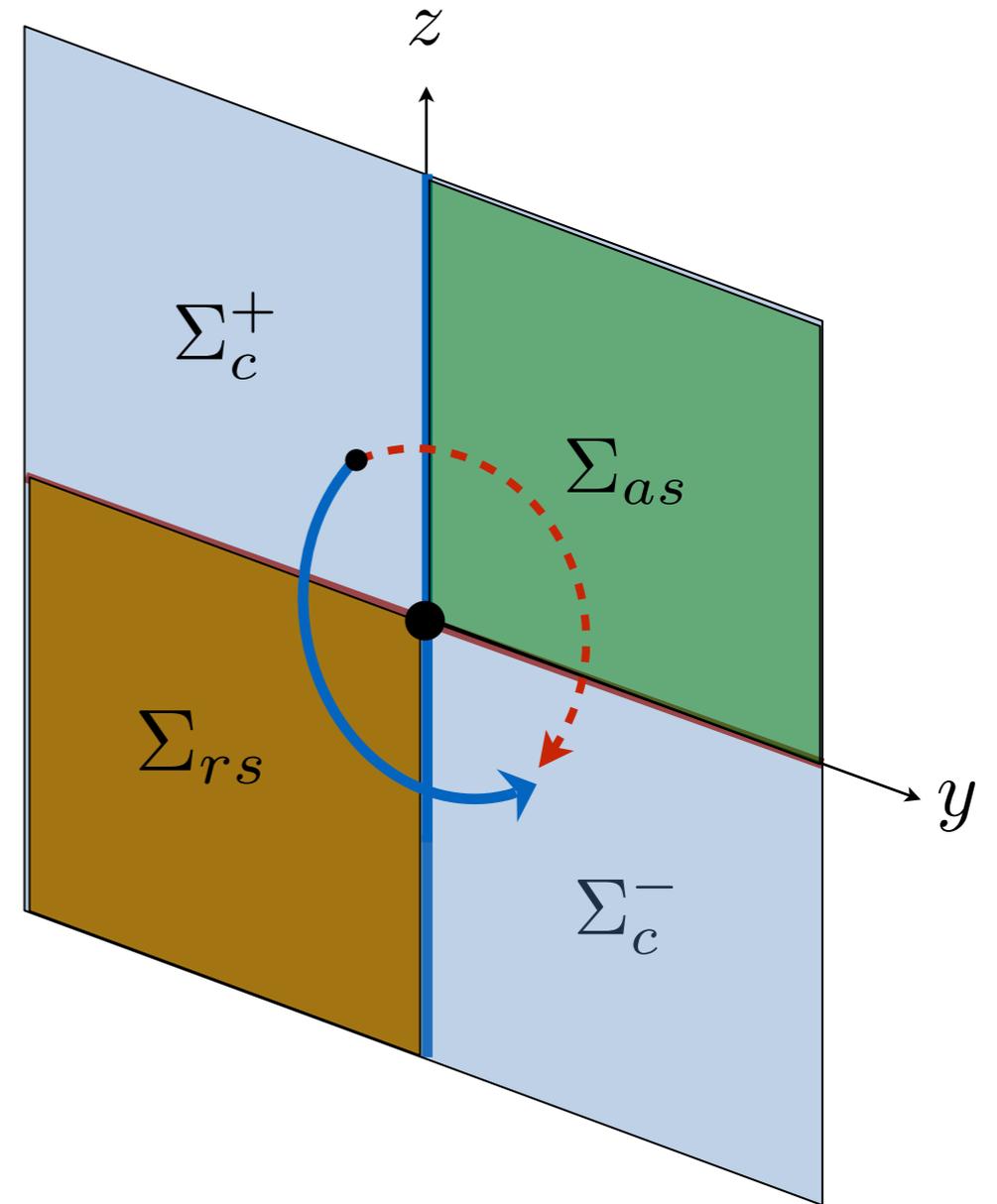
- (a) **(Supercritical case)** If $\kappa_S > 0$, then $\tilde{\mathbf{x}}(\varepsilon) \in \Sigma_{as}$ is a stable pseudo-node for $\varepsilon < 0$, being $\tilde{\mathbf{x}}(\varepsilon) \in \Sigma_{rs}$ a pseudo-saddle for $\varepsilon > 0$.
- (b) **(Subcritical case)** If $\kappa_S < 0$, then $\tilde{\mathbf{x}}(\varepsilon) \in \Sigma_{rs}$ is an unstable pseudo-node for $\varepsilon < 0$, being $\tilde{\mathbf{x}}(\varepsilon) \in \Sigma_{as}$ a pseudo-saddle for $\varepsilon > 0$.

$$\kappa_S = \begin{bmatrix} -v_+ & 1 \end{bmatrix} (v_+G - H) \begin{bmatrix} -v_+ \\ 1 \end{bmatrix}$$



The crossing dynamics bifurcation (periodic orbits?)

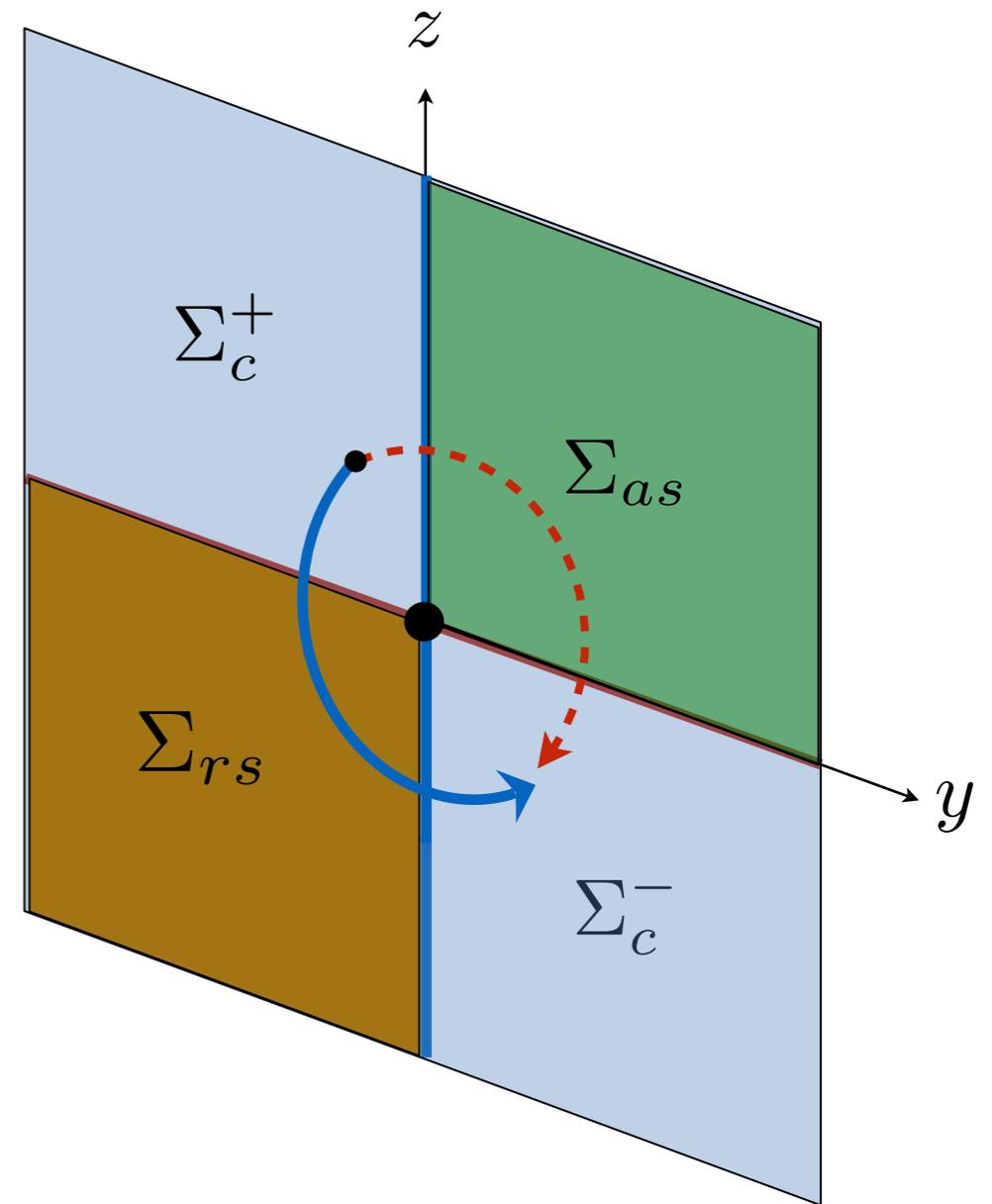
$$P_+ \begin{pmatrix} y \\ z \end{pmatrix} = P_-^{-1} \begin{pmatrix} y \\ z \end{pmatrix}$$



The crossing dynamics bifurcation (periodic orbits?)

$$P_+ \begin{pmatrix} y \\ z \end{pmatrix} = P_-^{-1} \begin{pmatrix} y \\ z \end{pmatrix} = P_- \begin{pmatrix} y \\ z \end{pmatrix}$$

Involution property



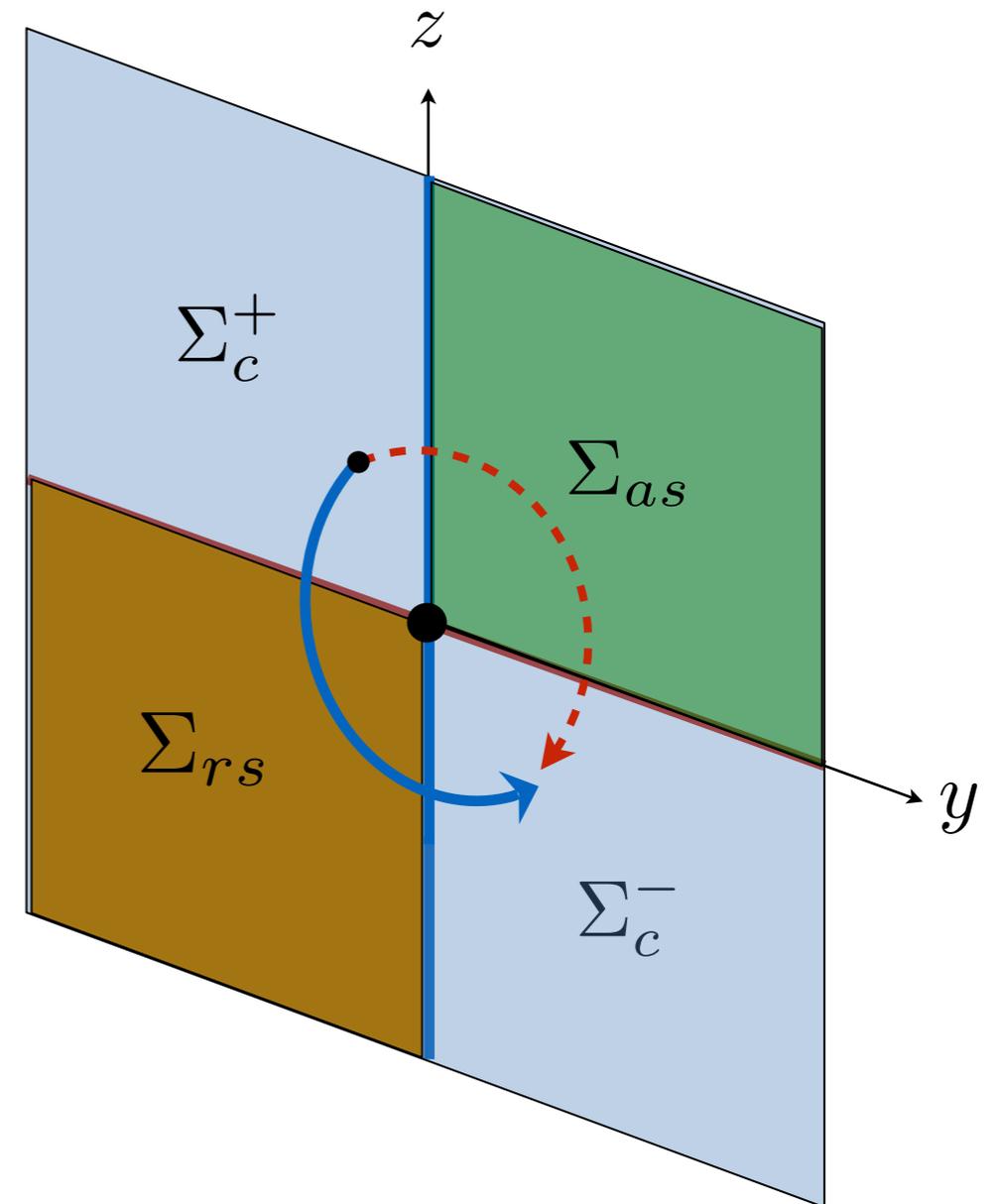
The crossing dynamics bifurcation (periodic orbits?)

$$P_+ \begin{pmatrix} y \\ z \end{pmatrix} = P_-^{-1} \begin{pmatrix} y \\ z \end{pmatrix} = P_- \begin{pmatrix} y \\ z \end{pmatrix}$$

Involution property

$$P_+ \begin{pmatrix} 0 \\ z \end{pmatrix} = \begin{pmatrix} 0 \\ z \end{pmatrix}; \quad P_- \begin{pmatrix} y \\ 0 \end{pmatrix} = \begin{pmatrix} y \\ 0 \end{pmatrix}$$

Invariance of the tangency lines



The structure of return maps

$$P_+ \begin{pmatrix} y \\ z \end{pmatrix} = \begin{bmatrix} -1 & 0 \\ -2v_+ & 1 \end{bmatrix} \begin{bmatrix} y \\ z \end{bmatrix} + y \begin{bmatrix} q_{11}^+ y + q_{12}^+ z \\ q_{21}^+ y + q_{22}^+ z \end{bmatrix} + y \begin{bmatrix} c_{11}^+ y^2 + c_{12}^+ yz + c_{13}^+ z^2 \\ c_{21}^+ y^2 + c_{22}^+ yz + c_{23}^+ z^2 \end{bmatrix} + O(4)$$

$$P_- \begin{pmatrix} y \\ z \end{pmatrix} = \begin{bmatrix} 1 & -2v_- \\ 0 & -1 \end{bmatrix} \begin{bmatrix} y \\ z \end{bmatrix} + z \begin{bmatrix} q_{11}^- y + q_{12}^- z \\ q_{21}^- y + q_{22}^- z \end{bmatrix} + z \begin{bmatrix} c_{11}^- y^2 + c_{12}^- yz + c_{13}^- z^2 \\ c_{21}^- y^2 + c_{22}^- yz + c_{23}^- z^2 \end{bmatrix} + O(4)$$

The structure of return maps

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$$\begin{aligned} q_{12}^+ &= 0; \\ q_{21}^+ &= v_+(q_{11}^+ - q_{22}^+); \\ c_{11}^+ &= -v_+c_{12}^+ - (q_{11}^+)^2; \\ c_{13}^+ &= 0; \\ c_{22}^+ &= v_+(c_{12}^+ - 2c_{23}^+) + q_{22}^+(q_{22}^+ - q_{11}^+)/2. \end{aligned}$$

Involution property

$$P_- \begin{pmatrix} y \\ z \end{pmatrix} = \begin{bmatrix} 1 & -2v_- \\ 0 & -1 \end{bmatrix} \begin{bmatrix} y \\ z \end{bmatrix} + z \begin{bmatrix} q_{11}^- y + q_{12}^- z \\ q_{21}^- y + q_{22}^- z \end{bmatrix} + z \begin{bmatrix} c_{11}^- y^2 + c_{12}^- yz + c_{13}^- z^2 \\ c_{21}^- y^2 + c_{22}^- yz + c_{23}^- z^2 \end{bmatrix} + O(4)$$

The structure of return maps

$$P_+ \begin{pmatrix} y \\ z \end{pmatrix} = \begin{bmatrix} -1 & 0 \\ -2v_+ & 1 \end{bmatrix} \begin{bmatrix} y \\ z \end{bmatrix} + y \begin{bmatrix} q_{11}^+ y + \cancel{q_{12}^+ z} \\ q_{21}^+ y + q_{22}^+ z \end{bmatrix} + y \begin{bmatrix} \cancel{c_{11}^+} y^2 + c_{12}^+ yz + \cancel{c_{13}^+} z^2 \\ c_{21}^+ y^2 + \cancel{c_{22}^+} yz + c_{23}^+ z^2 \end{bmatrix} + O(4)$$

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Involution property

$$P_- \begin{pmatrix} y \\ z \end{pmatrix} = \begin{bmatrix} 1 & -2v_- \\ 0 & -1 \end{bmatrix} \begin{bmatrix} y \\ z \end{bmatrix} + z \begin{bmatrix} q_{11}^- y + \cancel{q_{12}^-} z \\ \cancel{q_{21}^-} y + q_{22}^- z \end{bmatrix} + z \begin{bmatrix} c_{11}^- y^2 + \cancel{c_{12}^-} yz + c_{13}^- z^2 \\ \cancel{c_{21}^-} y^2 + c_{22}^- yz + \cancel{c_{23}^-} z^2 \end{bmatrix} + O(4)$$

$$\begin{aligned} q_{21}^- &= 0; \\ q_{12}^- &= v_-(q_{22}^- - q_{11}^-); \\ c_{23}^- &= -v_-c_{22}^- - (q_{22}^-)^2; \\ c_{21}^- &= 0; \\ c_{12}^- &= v_-(c_{22}^- - 2c_{11}^-) + q_{11}^-(q_{11}^- - q_{22}^-)/2. \end{aligned}$$

Involution property

Looking for non-trivial fixed points

We impose the equality $P_+(y, z) = P_-^{-1}(y, z)$, to obtain

$$\begin{aligned} 0 &= -2y + 2v_-z + q_{11}^+y^2 - q_{11}^-yz - q_{12}^-z^2 + c_{11}^+y^3 + (c_{12}^+ - c_{11}^-)y^2z - c_{12}^-yz^2 - c_{13}^-z^3 + O(4), \\ 0 &= -2v_+y + 2z + q_{21}^+y^2 + q_{22}^+yz - q_{22}^-z^2 + c_{21}^+y^3 + c_{22}^+y^2z + (c_{23}^+ - c_{22}^-)yz^2 - c_{23}^-z^3 + O(4). \end{aligned}$$



Again, we can have a bifurcation at $v_+v_- = 1$ so that we do a new bifurcation analysis by assuming all parameters fixed, excepting v_- , and take

$$v_- = \frac{1 + \varepsilon}{v_+}$$

Note that, from the second equation and the implicit function theorem, we can assume the existence of a function $p(y, \varepsilon)$ such that

$$z = y \cdot p(y, \varepsilon), \quad \text{with } p(0, 0) = v_+.$$

We need $v_+ < 0$

Looking for non-trivial fixed points

We get $z = y \cdot p(y, \varepsilon)$ with

$$p(y, \varepsilon) = v_+ - (q_{11}^+ - q_{22}^-(\varepsilon)v_+) \frac{y}{2} + [-2c_{21}^+ + q_{22}^+(2q_{11}^+ - q_{22}^+)v_+ + \\ + (2c_{22}^-(\varepsilon) - 2c_{12}^+ + 2c_{23}^+ - 2q_{22}^-(\varepsilon)q_{11}^+ - q_{22}^-(\varepsilon)q_{22}^+) v_+^2 - 2c_{22}^-(\varepsilon)v_-(\varepsilon)v_+^3] \frac{y^2}{4} + \dots$$

After substituting z in the first equation, we can desingularize it to get

$$0 = 2\varepsilon - (q_{11}^+ - q_{11}^-(\varepsilon)v_+) \varepsilon y + P(\varepsilon)y^2 + \dots$$

with

$$P(\varepsilon) = -\frac{1}{2v_+} [2c_{21}^+ + (2(q_{11}^+)^2 - 2q_{11}^+q_{22}^+ + (q_{22}^+)^2) v_+ + \\ + (-2c_{11}^-(\varepsilon) + 2c_{22}^-(\varepsilon) + 2c_{12}^+ - 2c_{23}^+ + q_{11}^-(\varepsilon)q_{11}^+ + q_{22}^-(\varepsilon)q_{22}^+) v_+^2 + \\ + (q_{11}^-(\varepsilon)^2 - 2q_{11}^-(\varepsilon)q_{22}^-(\varepsilon) + 2q_{22}^-(\varepsilon)^2) v_+^3 + 2c_{13}^-(\varepsilon)v_+^4 + O(\varepsilon)]$$

Looking for non-trivial fixed points

Clearly, it is more convenient to parameterize the emanating branch in terms of y , so that we get the expansion

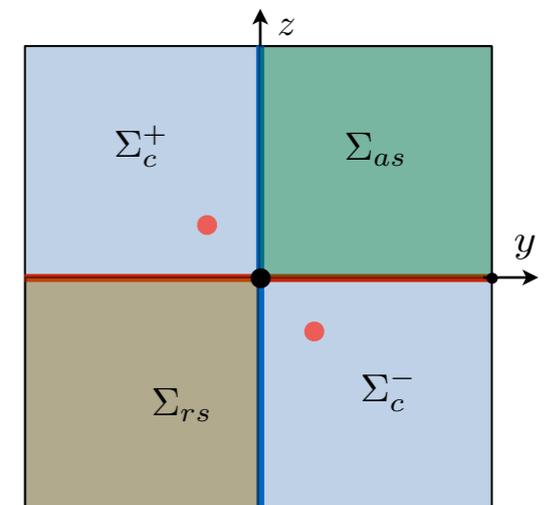
$$\varepsilon = \kappa_C \cdot y^2 + O(y^3),$$

where **the criticality coefficient is**

$$\begin{aligned} \kappa_C = & \frac{1}{4v_+} \left[2c_{21}^+ + (2(q_{11}^+)^2 - 2q_{11}^+q_{22}^+ + (q_{22}^+)^2) v_+ + \right. \\ & + \left(-2c_{11}^-(0) + 2c_{22}^-(0) + 2c_{12}^+ - 2c_{23}^+ + q_{11}^-(0)q_{11}^+ + q_{22}^-(0)q_{22}^+ \right) v_+^2 + \\ & \left. + (q_{11}^-(0)^2 - 2q_{11}^-(0)q_{22}^-(0) + 2q_{22}^-(0)^2) v_+^3 + 2c_{13}^-(0)v_+^4 \right] \end{aligned}$$

In short, depending on the sign of κ_C we have a subcritical or supercritical bifurcation of non-trivial fixed points with

$$\begin{aligned} z(y) &= v_+ \cdot y + O(y^2), \\ \varepsilon(y) &= \kappa_C \cdot y^2 + O(y^3). \end{aligned}$$



Topological type of non-trivial fixed points

Just computing the derivatives of the Poincaré half-return maps, and evaluating them at the branch of non-trivial fixed points, we get $DP(y) = D(P_- \circ P_+)(y, z(y))$ and the expansions for its determinant and trace, namely

$$\det DP(y) = 1 + d_1 y + d_2 y^2 + O(y^3),$$

$$\text{trace } DP(y) = 2 + t_1 y + t_2 y^2 + O(y^3).$$

It turns out that $t_1 = d_1$. When this common value vanishes, we have a degeneracy that should require much more long computations. We define $\sigma := t_1 = d_1$, and the computations give

$$\begin{aligned} \sigma &= -2q_{11}^+ + q_{22}^+ + (2q_{22}^-(0) - q_{11}^-(0)) v_+, \\ d_2 &= \frac{1}{2} \left[6(q_{11}^+)^2 - 5q_{11}^+ (q_{22}^+ - q_{11}^-(0)v_+ + 2q_{22}^-(0)v_+) + (q_{22}^+ - q_{11}^-(0)v_+ + 2q_{22}^-(0)v_+)^2 \right], \\ t_2 &= -\frac{1}{2v_+} \left[8c_{21}^+ + (2(q_{11}^+)^2 - 3q_{11}^+ q_{22}^+ + 3(q_{22}^+)^2) v_+ + \right. \\ &\quad \left. + (-8c_{11}^-(0) + 8c_{22}^-(0) + 8c_{12}^+ - 8c_{23}^+ - q_{11}^-(0)q_{11}^+ + 10q_{22}^-(0)q_{11}^+ + 2q_{11}^-(0)q_{22}^+) v_+^2 + \right. \\ &\quad \left. + (3q_{11}^-(0)^2 - 4q_{11}^-(0)q_{22}^-(0) + 4q_{22}^-(0)^2) v_+^3 + 8c_{13}^-(0)v_+^4 \right] \end{aligned}$$

Topological type of non-trivial fixed points

We have at the branch of non-trivial fixed points

$$\begin{aligned}\det DP(y) &= 1 + \sigma y + d_2 y^2 + O(y^3), \\ \text{trace } DP(y) &= 2 + \sigma y + t_2 y^2 + O(y^3).\end{aligned}$$

and, surprisingly, we get that $d_2 - t_2 = 8\kappa_C$.

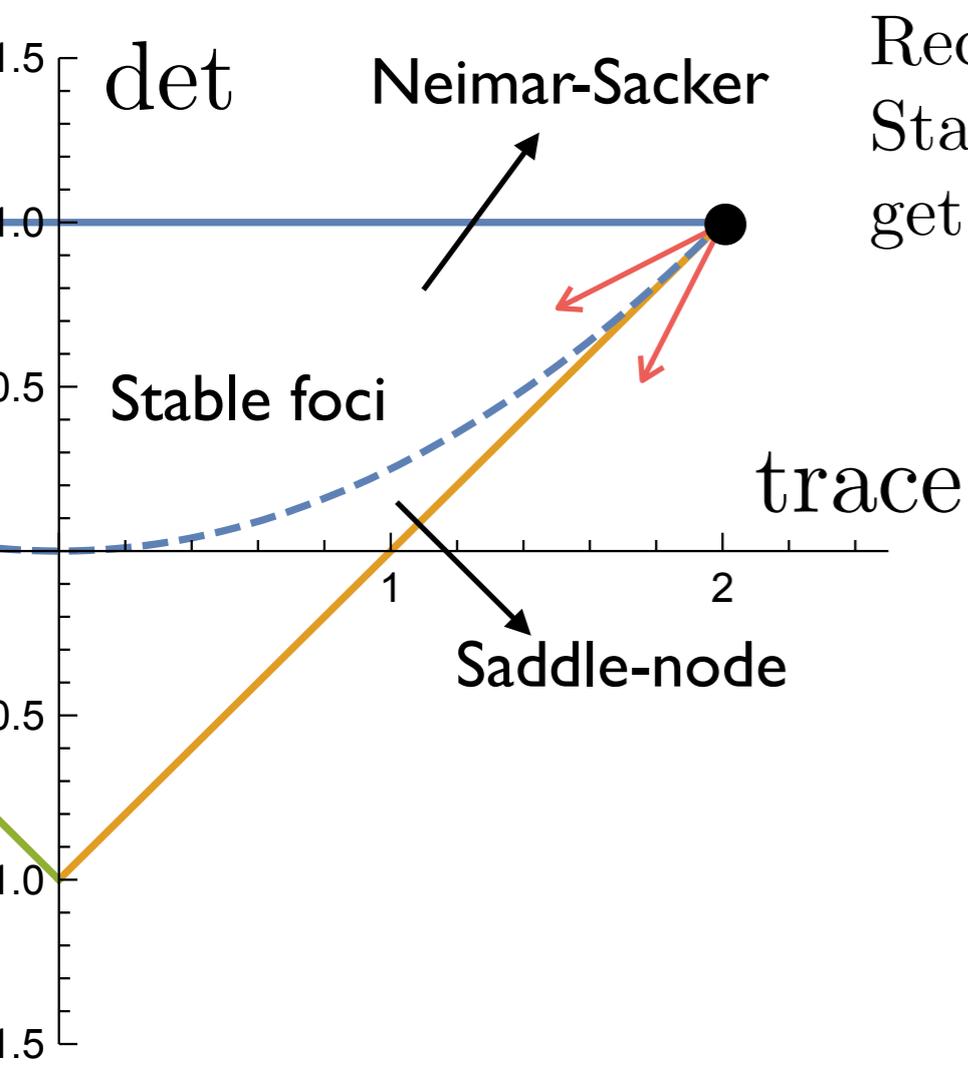
Topological type of non-trivial fixed points

We have at the branch of non-trivial fixed points

$$\det DP(y) = 1 + \sigma y + d_2 y^2 + O(y^3),$$

$$\text{trace } DP(y) = 2 + \sigma y + t_2 y^2 + O(y^3).$$

and, surprisingly, we get that $d_2 - t_2 = 8\kappa_C$.



Recall that we have $y < 0$ and small in absolute value. Stability requires $\sigma > 0$ and $\kappa_C > 0$. Furthermore, we get

$$\kappa_C < 0 \longrightarrow \text{SADDLE}$$

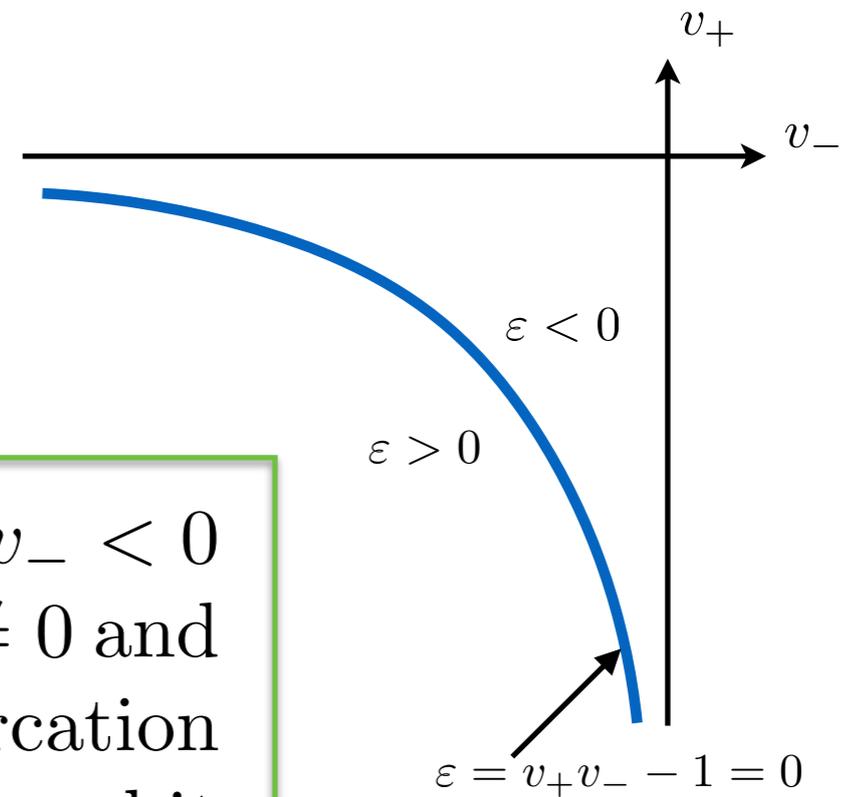
$$0 < \kappa_C < \frac{\sigma^2}{32} \longrightarrow \text{NODE}$$

$$\kappa_C > \frac{\sigma^2}{32} \longrightarrow \text{FOCUS}$$

Theorem C Assume in system (1) that $v_+ < 0$, $v_- < 0$ with $v_-v_+ - 1 = 0$ and that the two conditions $\sigma \neq 0$ and $\kappa_S \neq 0$ hold. By moving v_- and using the bifurcation parameter $\varepsilon = v_-v_+ - 1$, one crossing periodic orbit bifurcates from the origin for $\kappa_S \cdot \varepsilon > 0$ small.

The bifurcating periodic orbit is stable whenever $\sigma > 0$ and $\kappa_S > 0$.

The topological type of the corresponding fixed point for the Poincaré map is saddle, node or focus depending on whether $\kappa_S < 0$, $0 < \kappa_S < \sigma^2/32$ or $\kappa_S > \sigma^2/32$, respectively.



Computing the return maps

To compute the return map P_+ we write

$$\mathbf{x}(\tau) = e^{A^+ \tau} \mathbf{x}_0 + \int_0^\tau e^{A^+(\tau-s)} \mathbf{v}^+ ds,$$

where

$$A^+ = \begin{bmatrix} f_x^+ & -1 & 0 \\ g_x^+ & g_y^+ & g_z^+ \\ h_x^+ & h_y^+ & h_z^+ \end{bmatrix}, \quad \mathbf{x}_0 = \begin{bmatrix} 0 \\ y_0 \\ z_0 \end{bmatrix}, \quad \mathbf{v}^+ = \begin{bmatrix} 0 \\ 1 \\ v_+ \end{bmatrix},$$

being $y_0 < 0$. In practice, we take

$$\mathbf{x}(\tau) = \mathbf{x}_0 + \left(\tau I + \frac{\tau^2}{2} A^+ + \frac{\tau^3}{6} (A^+)^2 + \frac{\tau^4}{24} (A^+)^3 \right) M + \mathcal{O}(\tau^5)$$

where $M = A^+ \mathbf{x}_0 + \mathbf{v}^+$ and I is the identity matrix of order 3.

From the first component we can determine an expression for the time $\tau_+ = \tau_+(y_0, z_0)$ such that $x(\tau_+) = 0$. The third order polynomial approximation is given by

$$\tau_+(y_0, z_0) = a_{10}y_0 + a_{20}y_0^2 + a_{11}y_0z_0 + a_{30}y_0^3 + a_{21}y_0^2z_0 + a_{12}y_0z_0^2,$$

with

$$a_{10} = -2,$$

$$a_{20} = \frac{2}{3}(f_1^+ + g_2^+ - 2g_3^+v_+),$$

$$a_{11} = 2g_3^+,$$

$$a_{20} = \frac{2}{3}(f_1^+ + g_2^+ - 2g_3^+v_+),$$

$$a_{30} = -\frac{2}{9} \left(2(f_1^+ + g_2^+)^2 - 3(f_1^+g_2^+ - g_3^+h_2^+ + g_1^+) - (5f_1^+ + 5g_2^+ + 3h_3^+)g_3^+v_+ + 8(g_3^+v_+)^2 \right),$$

$$a_{12} = -2(g_3^+)^2,$$

$$a_{21} = \frac{4}{3}g_3^+(3g_3^+v_+ - f_1^+ - g_2^+ - h_3^+).$$

The half-return map $(y_1, z_1) = P_+(y_0, z_0)$ satisfies

$$y_1 = -y_0 + q_{11}^+ y_0^2 + y_0(c_{11}^+ y_0^2 + c_{12}^+ y_0 z_0) + O(4),$$

$$z_1 = -2v_+ y_0 + z_0 + y_0(q_{21}^+ y_0 + q_{22}^+ z_0) + y_0(c_{21}^+ y_0^2 + c_{22}^+ y_0 z_0 + c_{23}^+ z_0^2) + O(4),$$

where

$$\rightarrow q_{11}^+ = \frac{2}{3} (f_1^+ + g_2^+ + g_3^+ v_+),$$

$$\rightarrow q_{22}^+ = -2(h_3^+ - g_3^+ v_+),$$

$$q_{21}^+ = v_+(q_{11}^+ - q_{22}^+) = \frac{2}{3} (f_1^+ + g_2^+ - 2g_3^+ v_+ + 3h_3^+) v_+,$$

$$c_{11}^+ = -v_+ c_{12}^+ - (q_{11}^+)^2 = -\frac{2}{9} [2(f_1^+ + g_2^+)^2 + (f_1^+ + g_2^+ + 3h_3^+) g_3^+ v_+ - 4(g_3^+ v_+)^2],$$

$$\rightarrow c_{12}^+ = -\frac{2}{3} (f_1^+ + g_2^+ + 2g_3^+ v_+ - h_3^+) g_3^+,$$

$$\rightarrow c_{21}^+ = -\frac{2}{9} \{3(h_1^+ + f_1^+ h_2^+ - h_2^+ h_3^+) + [2(f_1^+ + g_2^+)^2 - 3(f_1^+ g_2^+ + g_1^+ - g_3^+ h_2^+) + 6h_3^+ (f_1^+ + g_2^+ + h_3^+)] v_+ + 5g_3^+ (f_1^+ + g_2^+ + 3h_3^+) v_+^2 + 8(g_3^+)^2 v_+^3\},$$

$$c_{22}^+ = v_+(c_{12}^+ - 2c_{23}^+) + q_{22}^+(q_{22}^+ - q_{11}^+)/2 =$$

$$= \frac{2}{3} [h_3^+ (f_1^+ + g_2^+ + 3h_3^+) - 2g_3^+ (f_1^+ + g_2^+ + 5h_3^+) v_+] + 4(g_3^+ v_+)^2,$$

$$\rightarrow c_{23}^+ = 2g_3^+ (h_3^+ - g_3^+ v_+) = -g_3^+ q_{22}^+.$$

Analogously, we can determine the half-return map $(y_2, z_2) = P_-^{-1}(y_0, z_0)$, getting

$$\begin{aligned} y_2 &= y_0 - 2v_- z_0 + z_0(q_{11}^- y_0 + q_{12}^- z_0) + z_0(c_{11}^- y_0^2 + c_{12}^- y_0 z_0 + c_{13}^- z_0^2) + O(4) \\ z_2 &= -z_0 + q_{22}^- z_0^2 + z_0(c_{22}^- y_0 z_0 + c_{23}^- z_0^2) + O(4), \end{aligned}$$

where

$$\rightarrow q_{11}^- = -2(g_2^- - h_2^- v_-),$$

$$q_{12}^- = v_-(q_{22}^- - q_{11}^-) = \frac{2}{3}(f_1^- + h_3^- - 2h_2^- v_- + 3g_2^-)v_-,$$

$$\rightarrow q_{22}^- = \frac{2}{3}(f_1^- + h_3^- + h_2^- v_-),$$

$$\rightarrow c_{11}^- = 2h_2^-(g_2^- - h_2^- v_-) = -h_2^- q_{11}^-,$$

$$\begin{aligned} c_{12}^- &= v_-(c_{22}^- - 2c_{11}^-) + q_{11}^-(q_{11}^- - q_{22}^-)/2 = \\ &= \frac{2}{3}[g_2^-(f_1^- + h_3^- + 3g_2^-) - 2h_2^-(f_1^- + h_3^- + 5g_2^-)v_-] + 4(h_2^- v_-)^2, \end{aligned}$$

$$\begin{aligned} \rightarrow c_{13}^- &= \frac{2}{9}\{3(g_1^- - f_1^- g_3^- + g_2^- g_3^-) - [2(f_1^- + h_3^-)^2 + 3(h_1^- + h_2^- g_3^- - f_1^- h_3^-) + \\ &+ 6g_2^-(f_1^- + g_2^- + h_3^-)] + 5h_2^-(f_1^- + 3g_2^- + h_3^-)v_-^2 - 8(h_2^-)^2 v_-^3\}, \end{aligned}$$

$$\rightarrow c_{22}^- = -\frac{2}{3}(f_1^- - g_2^- + 2h_2^- v_- + h_3^-)h_2^-,$$

$$c_{23}^- = -v_- c_{22}^- - (q_{22}^-)^2 = -\frac{2}{9}[2(f_1^- + h_3^-)^2 + h_2^-(f_1^- + 3g_2^- + h_3^-)v_- - 4(h_2^- v_-)^2].$$

Examples & Applications

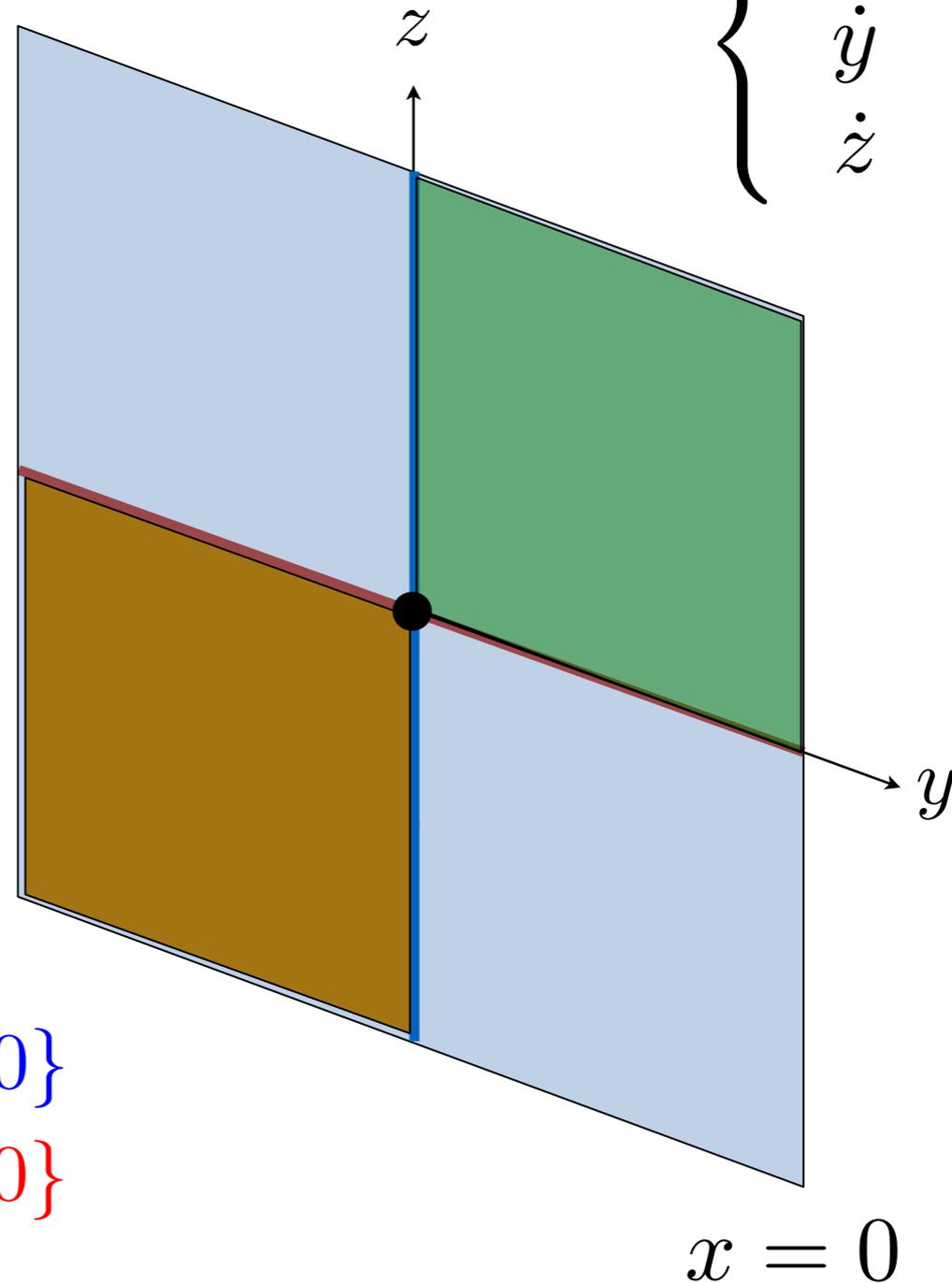
A multi-parametric example

$$\begin{cases} \dot{x} &= -y \\ \dot{y} &= 1 \\ \dot{z} &= ay + bz + v_+ \end{cases}$$

$$a < 0$$

$$b < 0$$

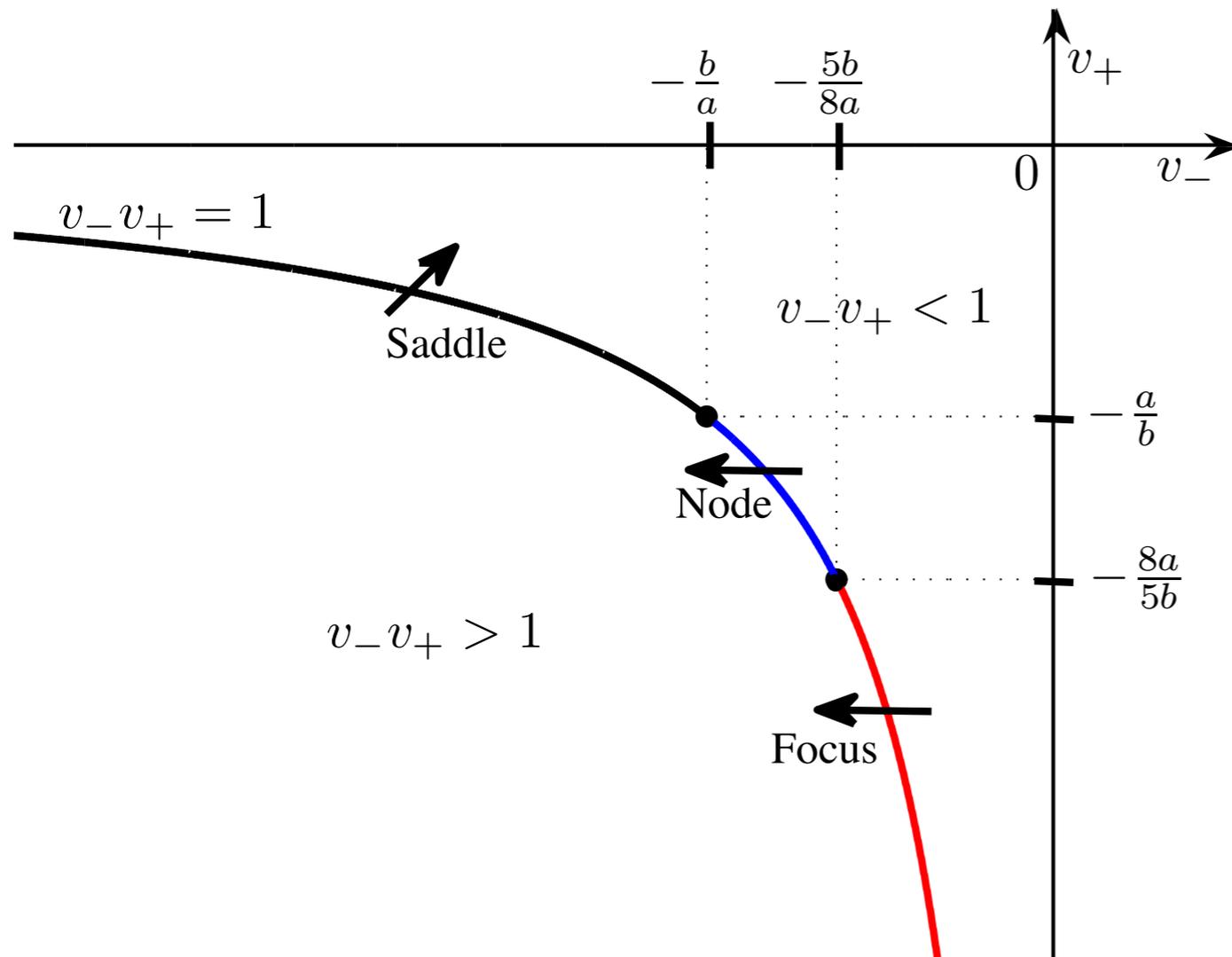
$$\begin{cases} \dot{x} &= z \\ \dot{y} &= v_- \\ \dot{z} &= 1 \end{cases}$$



$$T_+ = \{(x, y, z) \in \mathbb{R}^3 : x = y = 0\}$$

$$T_- = \{(x, y, z) \in \mathbb{R}^3 : x = z = 0\}$$

A multi-parametric example



We have for this example

$$\kappa_S = av_+ - b,$$

$$\sigma = -2b,$$

$$\kappa_C = b \frac{a + bv_+}{3v_+}.$$

$$a < 0$$

$$b < 0$$

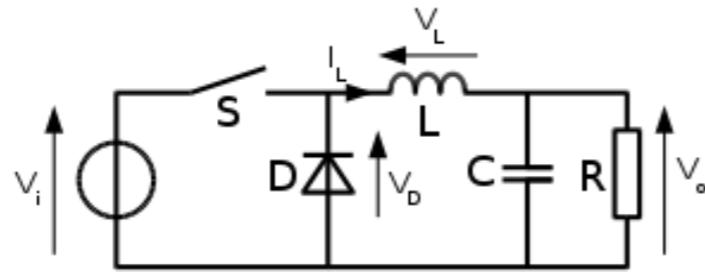
Bifurcation set to be completed.

See the preprint by A. Algaba, E. Freire, E. Gamero and C. García,
Bifurcation analysis of planar nilpotent reversible systems.

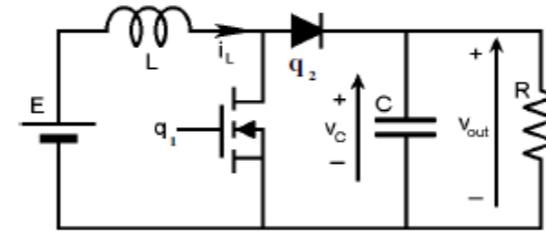
Electronic converters

- Electronic Power converters are **Switching-Mode Power Supplies** (SMPS).
- Basically, they are built using **semiconductor switches** (diodes, transistors) and **energy storage elements** (inductors, capacitors)
- **Examples**: Rectifiers AC-DC, inverters DC-AC, **DC-DC converters** (buck, boost, buck-boost)
- Their mathematical models commonly lead to **non-smooth** dynamical systems

DC-DC converters

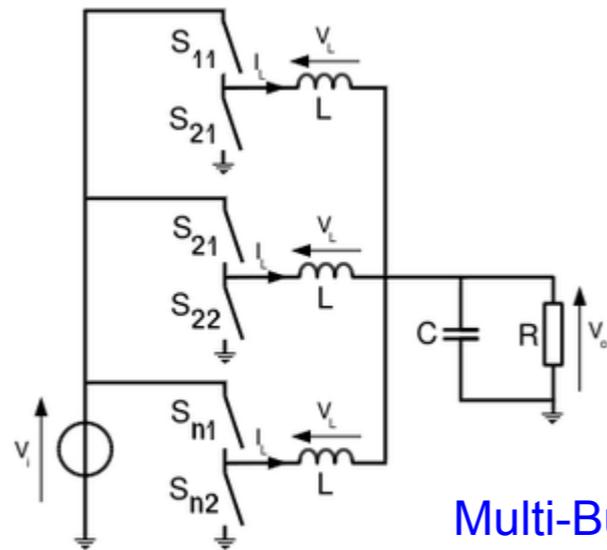
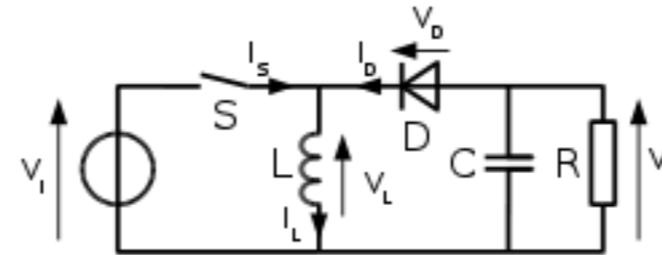


Buck (step-down)

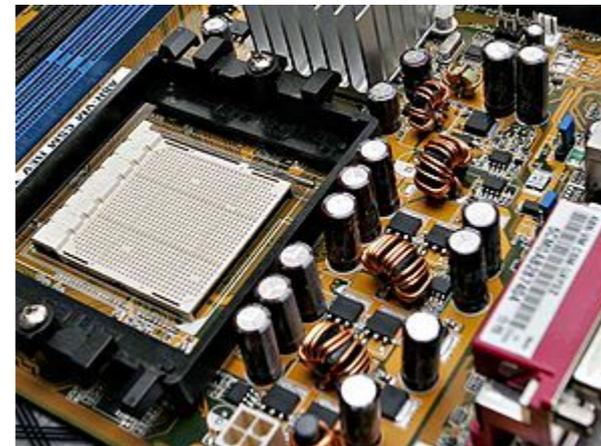


Boost (step-up)

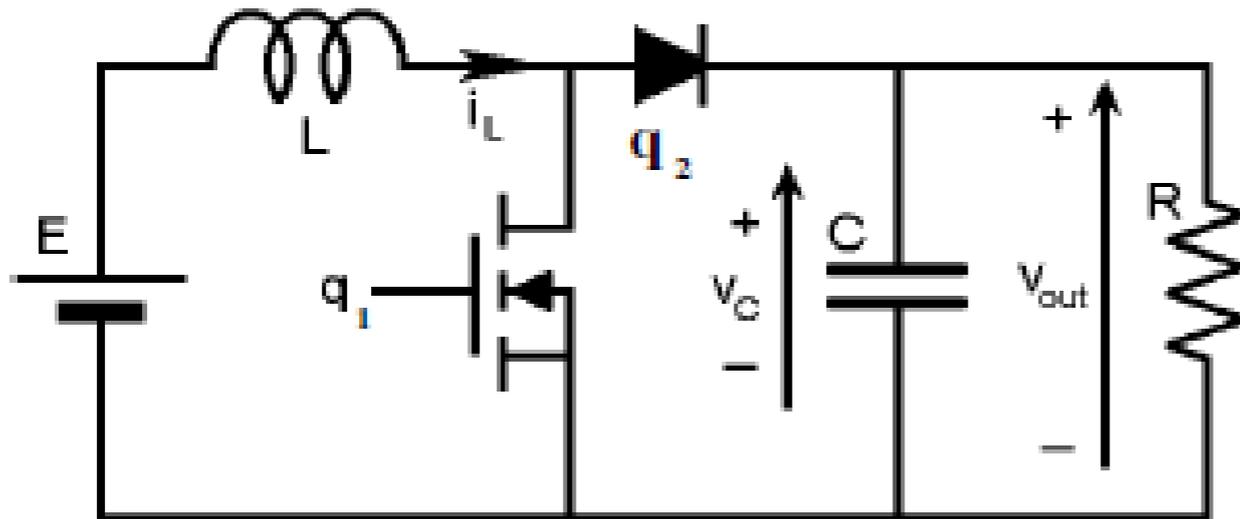
Buck-Boost (step-up or step-down)



Multi-Buck (CPU)



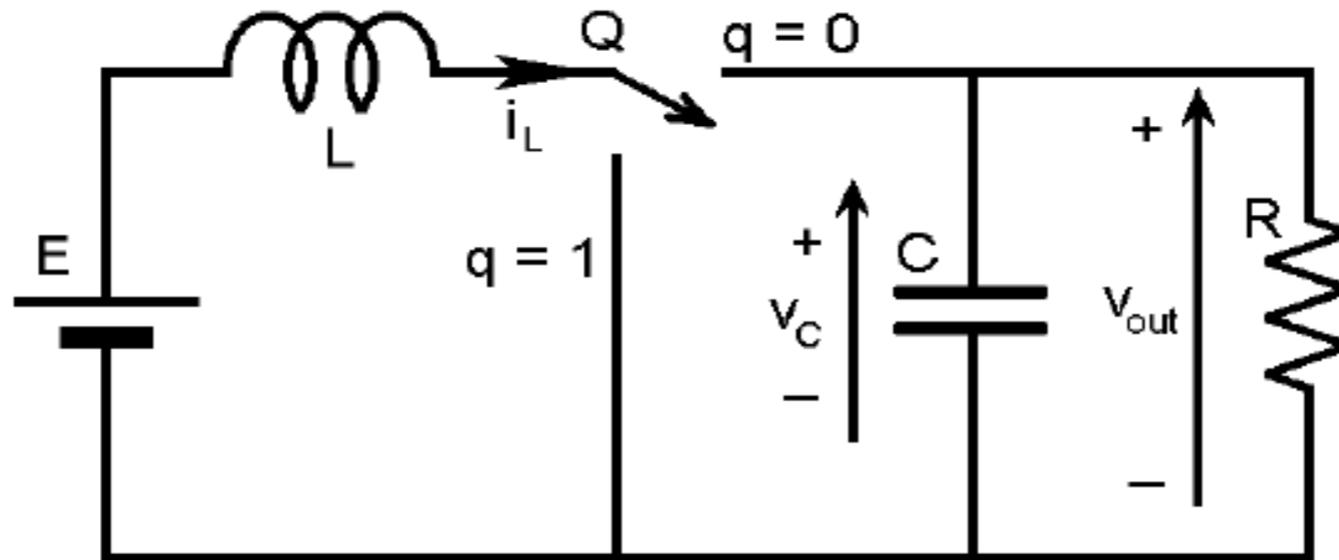
The BOOST converter



- The value of q_1 stands for a controlled switch
- when q_1 is turned **ON** → current in L increases and energy is stored in it
- when q_1 is turned **OFF** → the stored energy in L is dropped and the polarity of the L voltage changes so that it adds to the input voltage

The goal is to get $V_{out} > E$

The BOOST converter



The model of the system is given by

$$\begin{aligned} L \frac{di_L}{dt} &= V_{in} - r_L i_L - u \cdot v_c \\ C \frac{dv_c}{dt} &= u \cdot i_L - \frac{v_c}{R} \end{aligned} \quad (V_{in} = E)$$

where $v_c > 0$ is the capacitor voltage, $i_L > 0$ is the inductor current and $u \in \{0, 1\}$ is the control action. Input voltage is assigned as V_{in} , r_L is the equivalent series resistance of the inductor, R is the resistive load, C and L are the capacitor and inductor, respectively.

The BOOST converter

To analyze the model, the system is normalized

$$(i_L, v_c) = \left(V_{in} \sqrt{\frac{C}{L}} x, V_{in} y \right) \quad \text{and} \quad t = \tau \sqrt{LC};$$

and new parameters are taken, namely

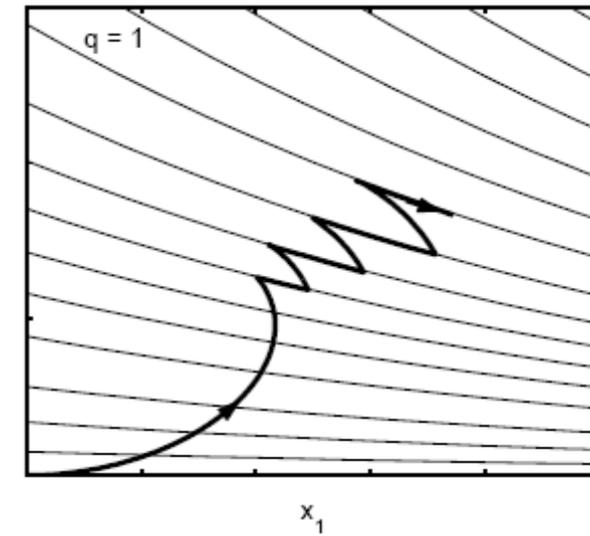
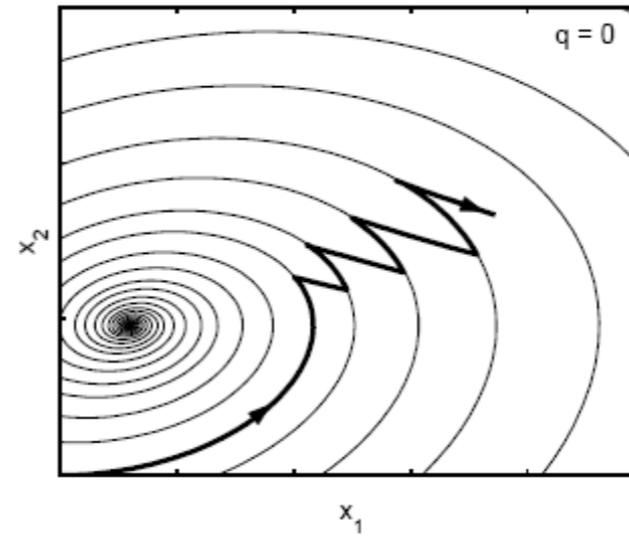
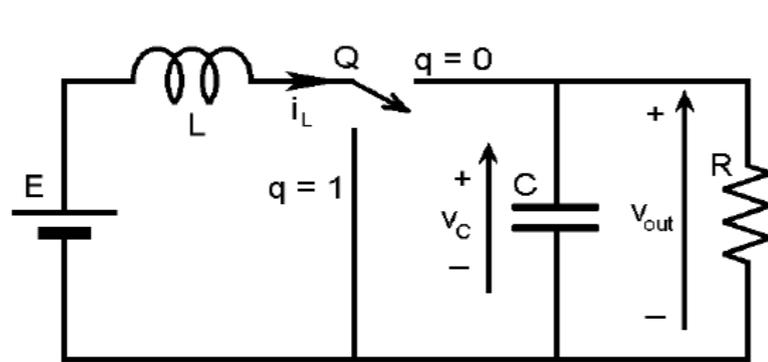
$$b = r_L \sqrt{\frac{C}{L}} \quad \text{and} \quad a = \frac{1}{R} \sqrt{\frac{L}{C}};$$

so that the boost converter model in dimensionless normal form is

$$\begin{aligned} \dot{x} &= 1 - bx - u \cdot y \\ \dot{y} &= u \cdot x - ay, \end{aligned}$$

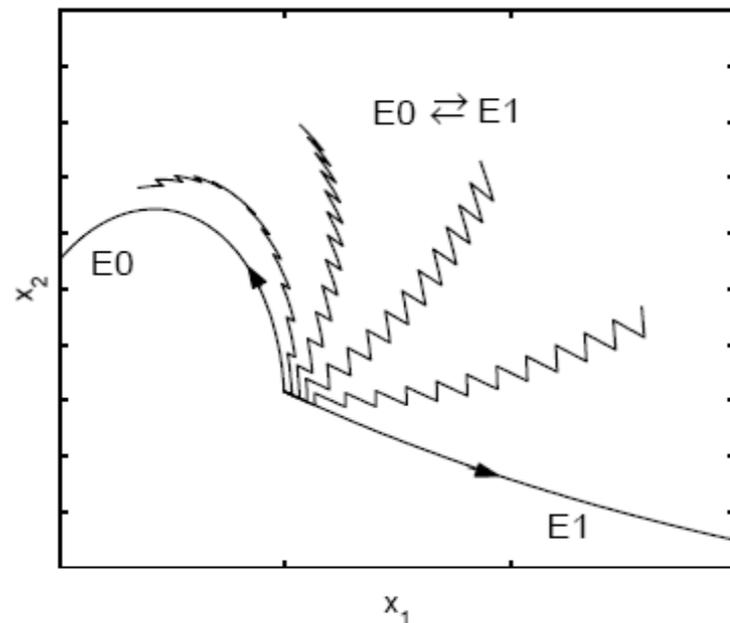
for $x, y > 0$, $u = \{0, 1\}$, $b \geq 0$, and $a > 0$.

The strategy at the BOOST converter



$$E_0 \begin{cases} \dot{x} = 1 - bx - y \\ \dot{y} = x - ay \end{cases}$$

$$E_1 \begin{cases} \dot{x} = 1 - bx \\ \dot{y} = -ay \end{cases}$$



- Frequency variable Control
- Frequency constant Control using **PWM** (Pulse-Width Modulation)

Control of the Boost converter

We want to regulate the output voltage to a desired value $y = v_c/V_{in} = y_r > 1$, ensuring robustness under parameter variation of a , produced by load changes in R .

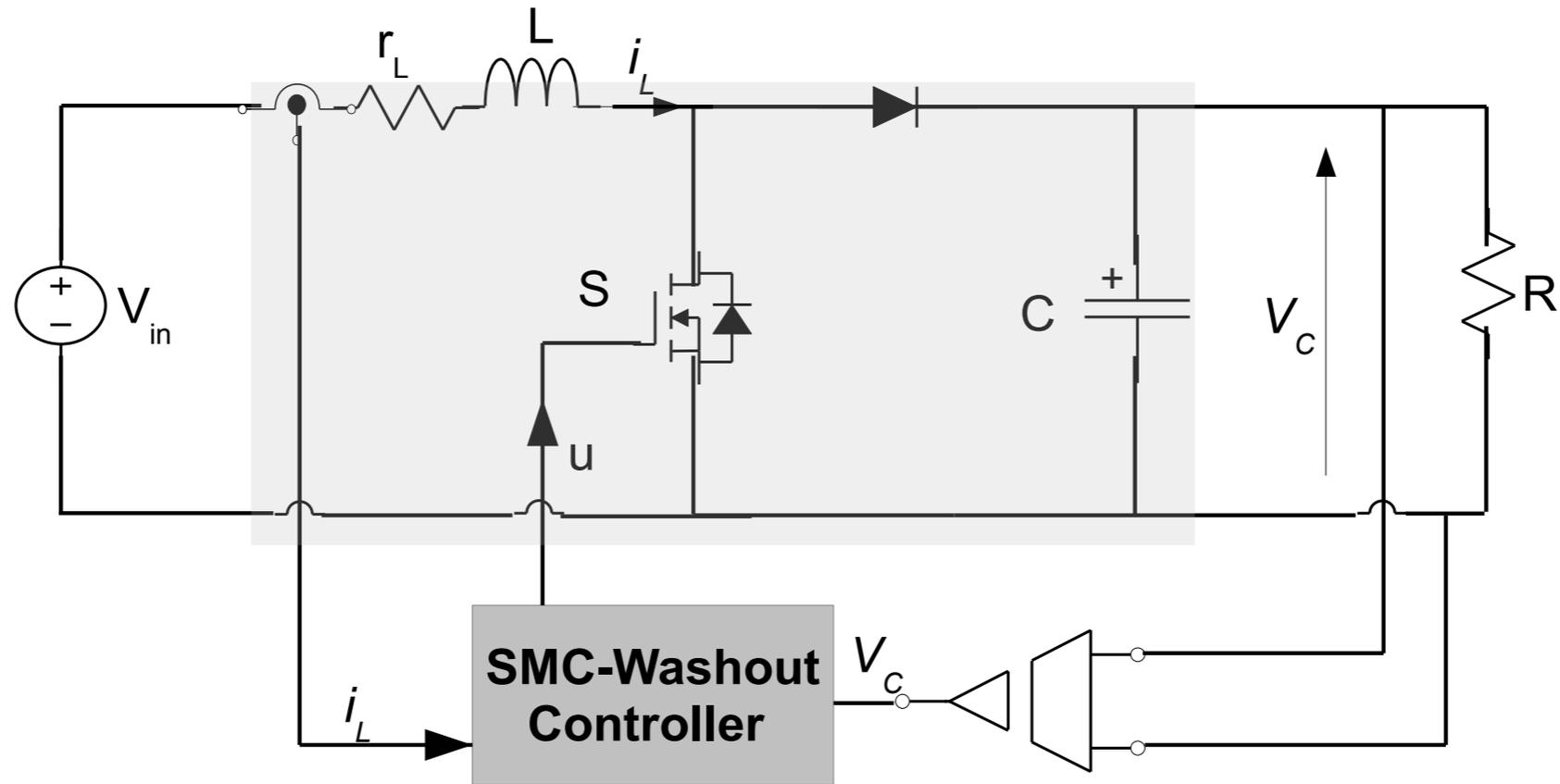
A washout filter is used: the inductor current x can be filtered to get a new signal x_F by means of a washout filter given by the transfer function

$$G_F(s) = \frac{X_F(s)}{X(s)} = \frac{s}{s + w} = 1 - \frac{w}{s + w},$$

where w is the reciprocal of the filter constant and x_F is the filter output.

A differential equation is added, $\dot{z} = w(x - z)$, where z is a new state satisfying the output equation $x_F = x - z$.

Control of the Boost converter



Control of the Boost converter

The SMC strategy consists in the choice of

$$\Sigma = \{\mathbf{x} = (x, y, z) \in \mathbb{R}^3 : h(\mathbf{x}) = y - y_r + k(x - z) = 0\}$$

as the switching boundary, where we want to be located the pseudo-equilibrium point.

We use the vector fields defined by

$$\mathbf{F}^+(\mathbf{x}) = \begin{bmatrix} 1 - bx - y \\ x - ay \\ w(x - z) \end{bmatrix} \quad \text{and} \quad \mathbf{F}^-(\mathbf{x}) = \begin{bmatrix} 1 - bx \\ -ay \\ w(x - z) \end{bmatrix},$$

for $h(\mathbf{x}) > 0$ and $h(\mathbf{x}) < 0$, respectively.

In what follows, we assume for simplicity $b = 0$ (ideal inductance), $w = 1$ and $y_r = 2$ (doubling the voltage).

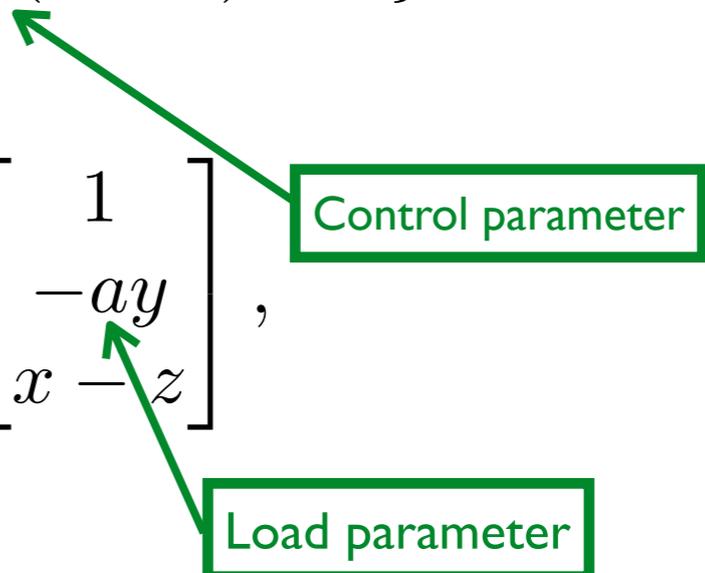
Control parameter

Control of the Boost converter

Therefore, we work with

$$\Sigma = \{\mathbf{x} = (x, y, z) \in \mathbb{R}^3 : h(\mathbf{x}) = y - 2 + k(x - z) = 0\}$$

and

$$\mathbf{F}^+(\mathbf{x}) = \begin{bmatrix} 1 - y \\ x - ay \\ x - z \end{bmatrix} \quad \text{and} \quad \mathbf{F}^-(\mathbf{x}) = \begin{bmatrix} 1 \\ -ay \\ x - z \end{bmatrix},$$


for $h(\mathbf{x}) > 0$ and $h(\mathbf{x}) < 0$, respectively.

We need to compute $\nabla h \cdot \mathbf{F}^\pm$ on Σ , and look for Σ_{as} , namely

$$\nabla h \cdot \mathbf{F}^+(\mathbf{x})|_{\Sigma} = [k, 1, -k] \begin{bmatrix} 1 - y \\ x - ay \\ x - z \end{bmatrix} = k(1 - y) + x - ay - k(x - z) = x + (1 - a - k)y + k - 2 < 0,$$

and

$$\nabla h \cdot \mathbf{F}^-(\mathbf{x})|_{\Sigma} = [k, 1, -k] \begin{bmatrix} 1 \\ -ay \\ x - z \end{bmatrix} = k - ay - k(x - z) = (1 - a)y + k - 2 > 0.$$

TS-point in a DC-DC Boost converter

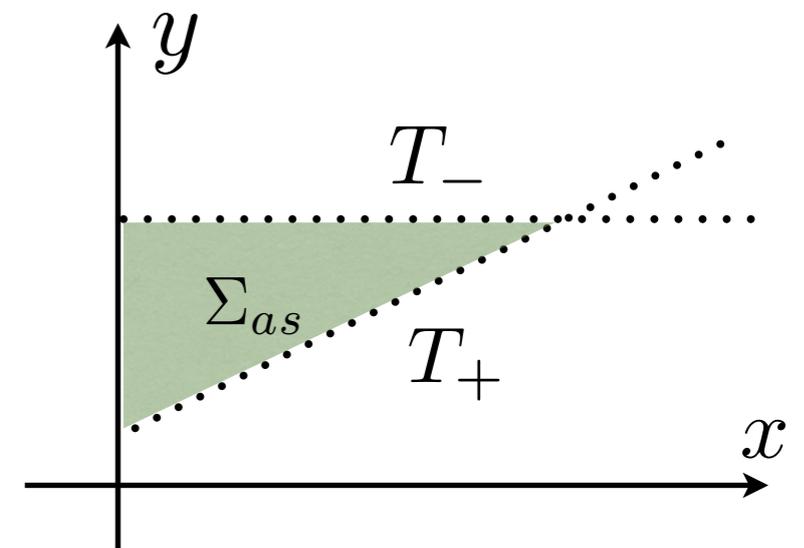
The tangency lines are

$$T_- = \{\mathbf{x} \in \Sigma : (1 - a)y + k - 2 = 0\},$$

$$T_+ = \{\mathbf{x} \in \Sigma : x + (1 - a - k)y + k - 2 = 0\}.$$

The double tangency point occurs where the tangency lines intersect transversally, i.e., at the point $\hat{\mathbf{x}} = (k\hat{y}, \hat{y}, \hat{z})$, with

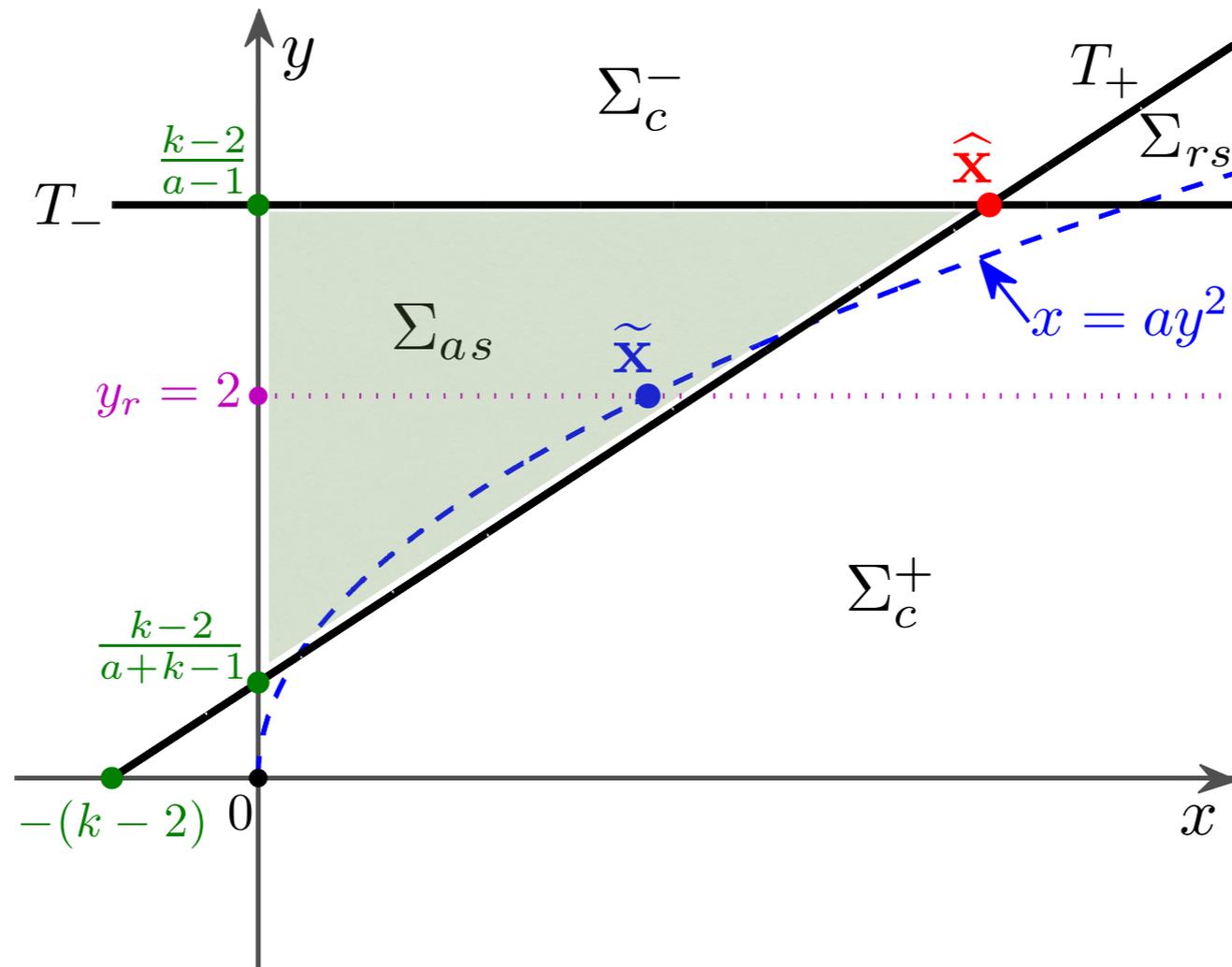
$$\hat{y} = \frac{k - 2}{a - 1},$$
$$\hat{z} = \frac{(1 + k^2)\hat{y} - 2}{k}.$$



where we assume $a \neq 1$.

To have $\hat{y} > 0$, we will consider $\text{sign}(k - 2) = \text{sign}(a - 1)$.

TS-point in a DC-DC Boost converter



TS-point in a DC-DC Boost converter

The pseudo-equilibrium point is at $(\tilde{x}, \tilde{y}, \tilde{z}) = (4a, 2, 4a)$ to be in Σ_{as} only if $k > 2a$. Thus, we must expect to have the compound bifurcation for $k = 2a$.

First, we need to check the conditions **(H2)** to have a TS-point. Computations lead to the two conditions

$$k > 2, \quad 1 < a < \frac{k - 3 + \sqrt{17 - 34k + 37k^2 - 20k^3 + 4k^4}}{2(k - 2)}.$$

Assuming these inequalities, we put the system in the canonical form and compute the critical coefficients.

We obtain for the new coefficients

$$v_- = \frac{1 - a^2(k - 2) - a(3 - 3k + k^2)}{(a - 1)\omega_+\omega_-}, \quad v_+ = \frac{(a - k)(k - 2)}{\omega_+\omega_-},$$

along with

$$\begin{aligned} g_2^- &= h_1^- = h_2^- = 0, & g_1^- &= -\frac{1}{k\omega_-\omega_+}, & f_1^- &= -\frac{1}{\omega_-}, \\ g_3^- &= \frac{a^2k + a(1 - k + k^2) - 1}{(a - 1)k\omega_+}, & h_3^- &= -\frac{a}{\omega_-}, & f_1^+ &= -\frac{k - 1}{k\omega_+}, \\ g_1^+ &= \frac{k^2 + (a - 2)k + 1}{k^2\omega_+^2}, & g_2^+ &= -\frac{k^2 + (a - 1)k + 1}{k\omega_+}, & h_1^+ &= -\frac{a - 1}{k\omega_-\omega_+}, \\ g_3^+ &= -\frac{(k^2 + (a - 2)k + 2 - a)\omega_-}{(a - 1)\omega_+^2}, & h_2^+ &= \frac{a - 1}{\omega_-}, & h_3^+ &= \frac{k - 1}{\omega_+}, \end{aligned}$$

where the two conditions

$$\omega_-^2 = a(k - 2) > 0, \quad \omega_+^2 = \frac{(k - 1)^3 + (k - 3)a - (k - 2)a^2}{a - 1} > 0,$$

are guaranteed from Hypothesis (H2).

TS-point in a DC-DC Boost converter

The bifurcation curve in the (a, k) -plane of parameters turns out to be

$$v_- v_+ - 1 = (2a - k)\hat{y} = 0,$$

i.e.,

$$k = 2a.$$

Moreover, we have

$$v_+ \Big|_{k=2a} = \frac{2a(1-a)}{\omega_+ \omega_-} < 0 \quad \text{and} \quad v_- \Big|_{k=2a} = -\frac{(1-a)^2 + 5a^2}{\omega_+ \omega_-} < 0$$

for all $a > 1$. The sliding bifurcation is supercritical, since

$$\kappa_S \Big|_{k=2a} = \frac{2}{((1-a)^2 + 5a^2)^{\frac{3}{2}}} > 0.$$

TS-point in a DC-DC Boost converter

Regarding the crossing bifurcation, we obtain

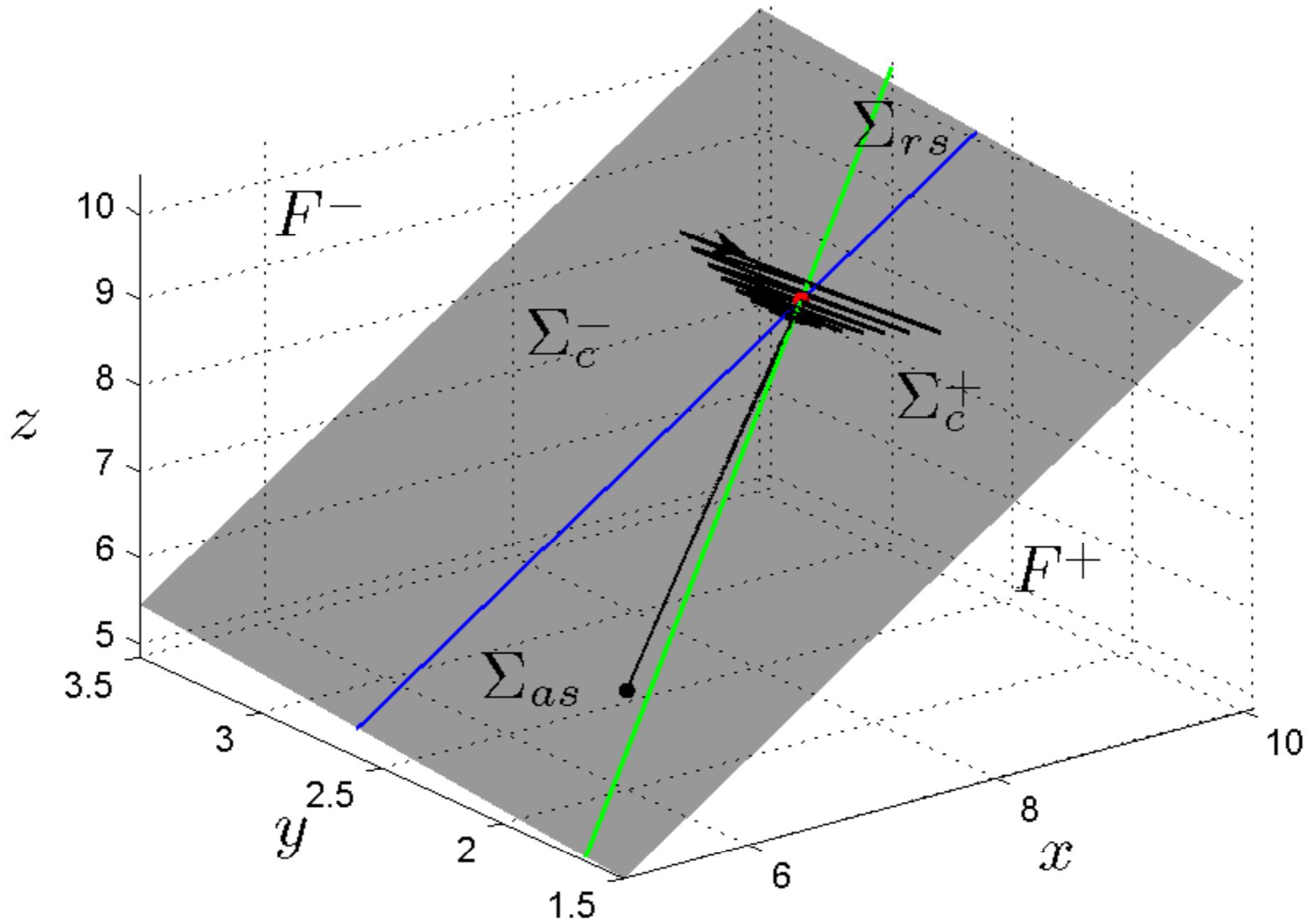
$$\sigma|_{k=2a} = \frac{2(5 - 4a + 16a^2 + 24a^3)}{3(1 - 2a + 6a^2)^{\frac{3}{2}}} > 0,$$

and

$$\kappa_C = \frac{2}{3((1 - a)^2 + 5a^2)^2} > 0,$$

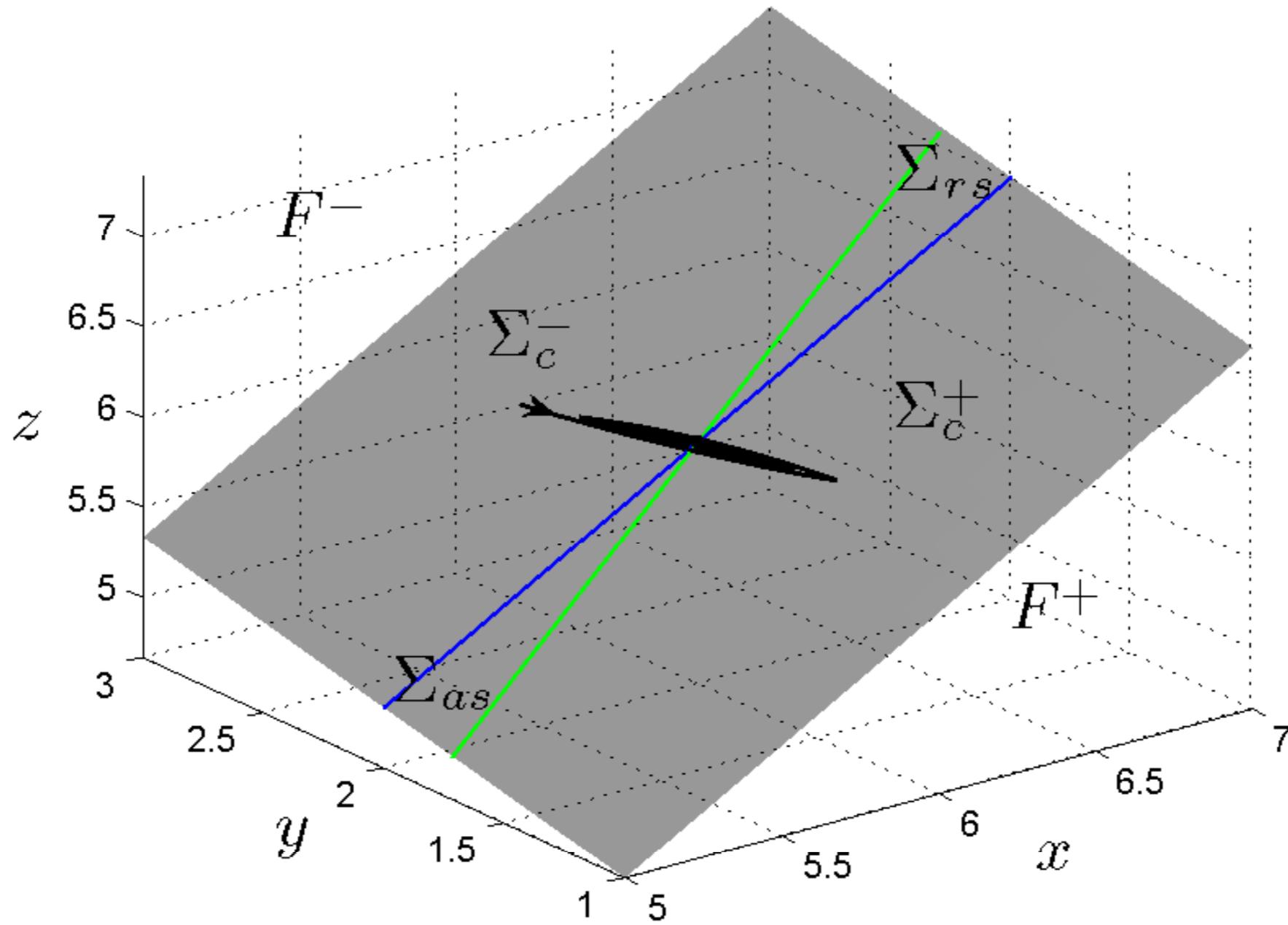
so that the bifurcating periodic orbit for $k < 2a$ is of **stable node** type.

Simulation results



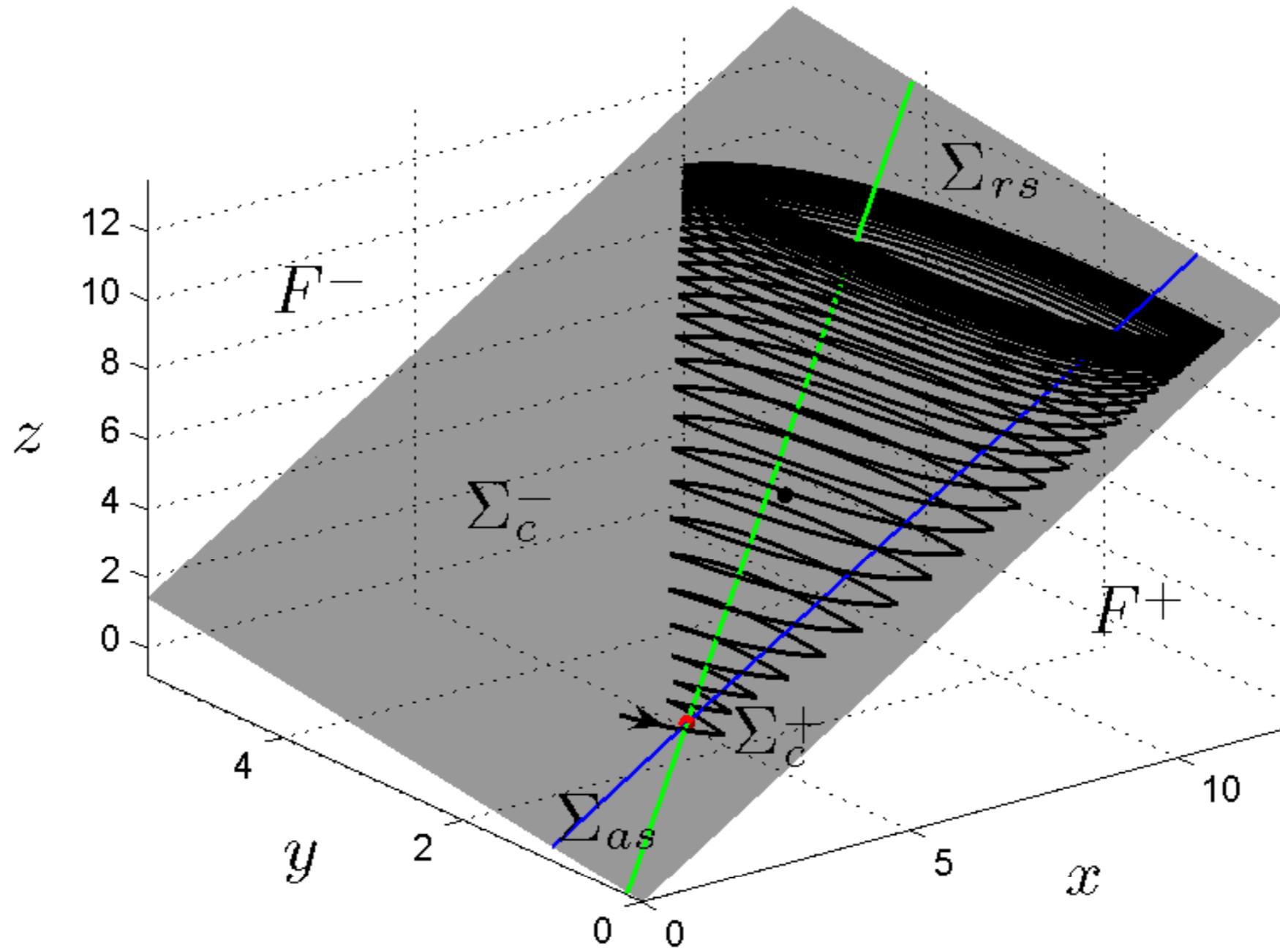
$$3.3 = k > 2a = 3$$

Simulation results



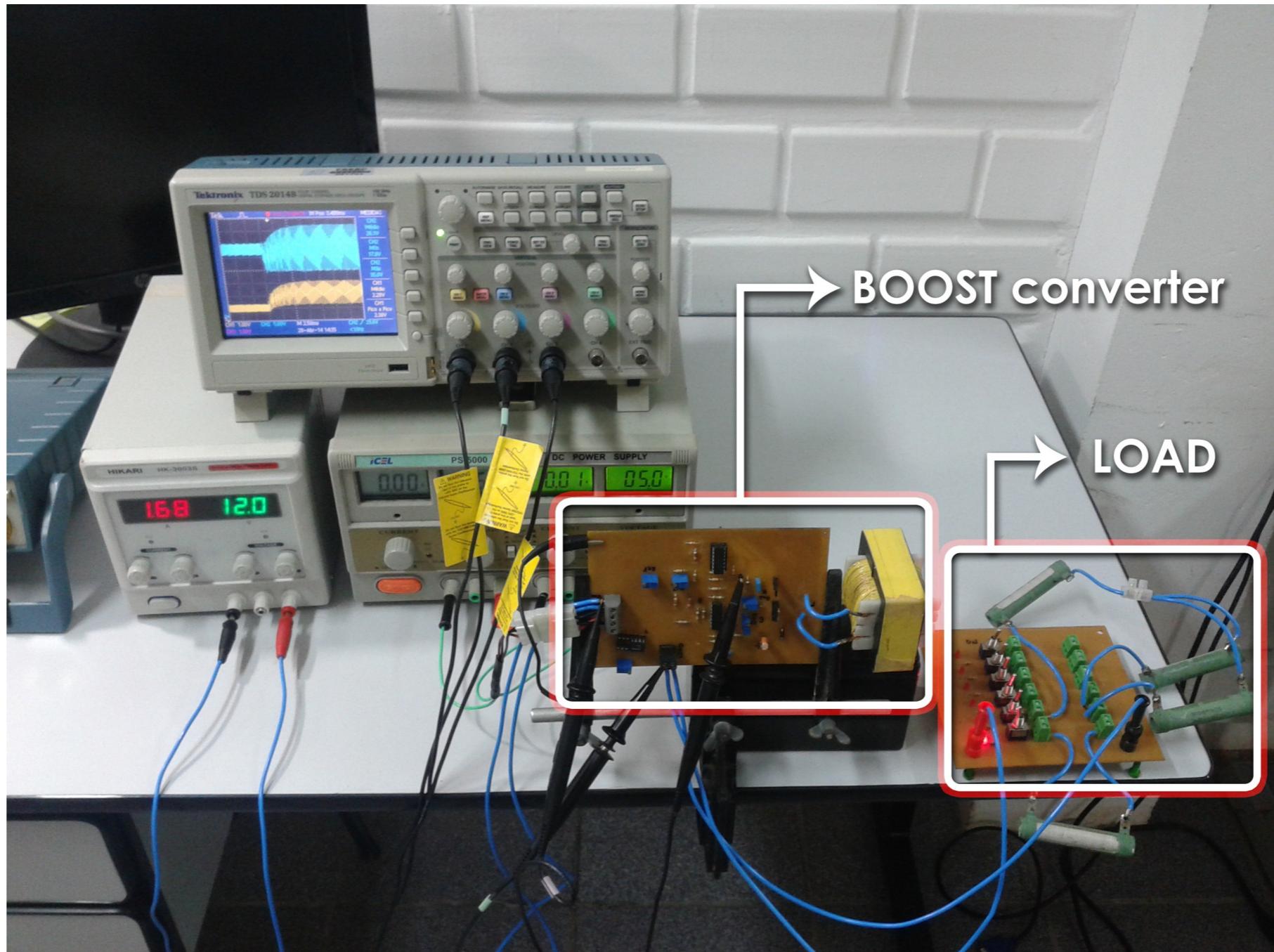
$$k = 2a = 3$$

Simulation results



$$2.5 = k < 2a = 3$$

Undesirable oscillation observed in laboratory due to the TS-point in the DC-DC Boost converter



Conclusions

- The TS compound bifurcation has been characterized for piecewise linear systems, and a procedure for computing the essential coefficients has been provided.
- Several examples have been shown, and in particular the bifurcation is detected in DC-DC converters under SMC strategy with a washout filter.
- A codimension-two unfolding is still needed, gaining information on secondary bifurcation curves.
- A question to solve: Can this study with linear pieces serve as a normal form for nonlinear vector fields?

Thank you for your attention!